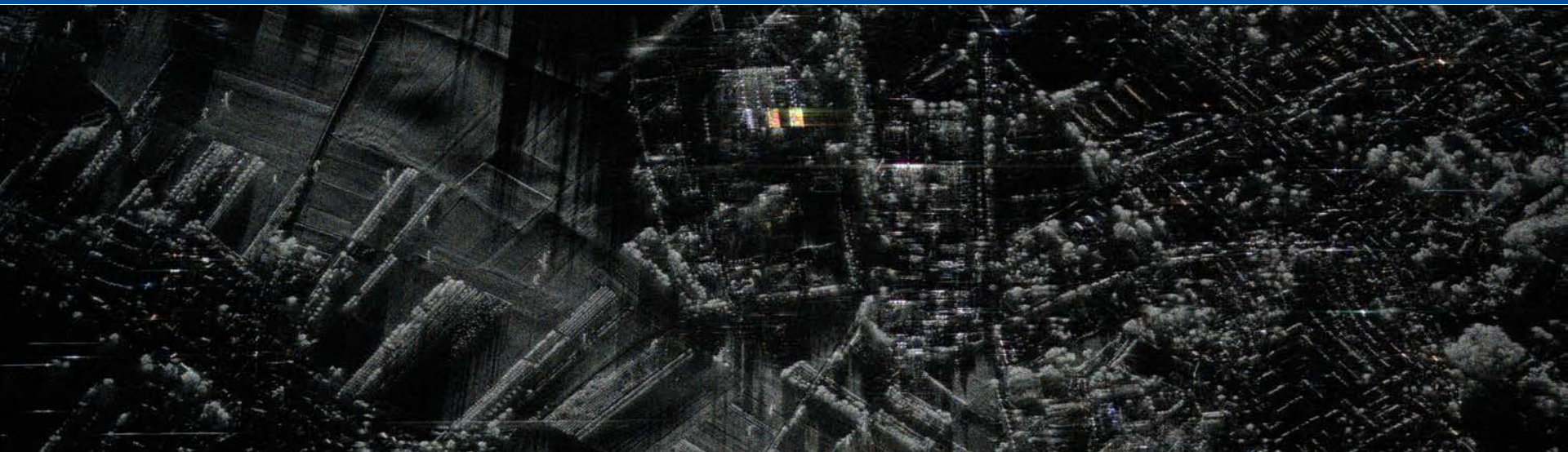


Passive Radar at ZESS – From HITCHHIKER to ASTRA

Otmar Loffeld, Holger Nies, Florian Behner, Simon Reuter
University of Siegen, Center for Sensor Systems



Overview of the project

HITCHHIKER



Overview of the project

HITCHHIKER

First Spaceborne-Stationary
Experiment with TerraSAR-X

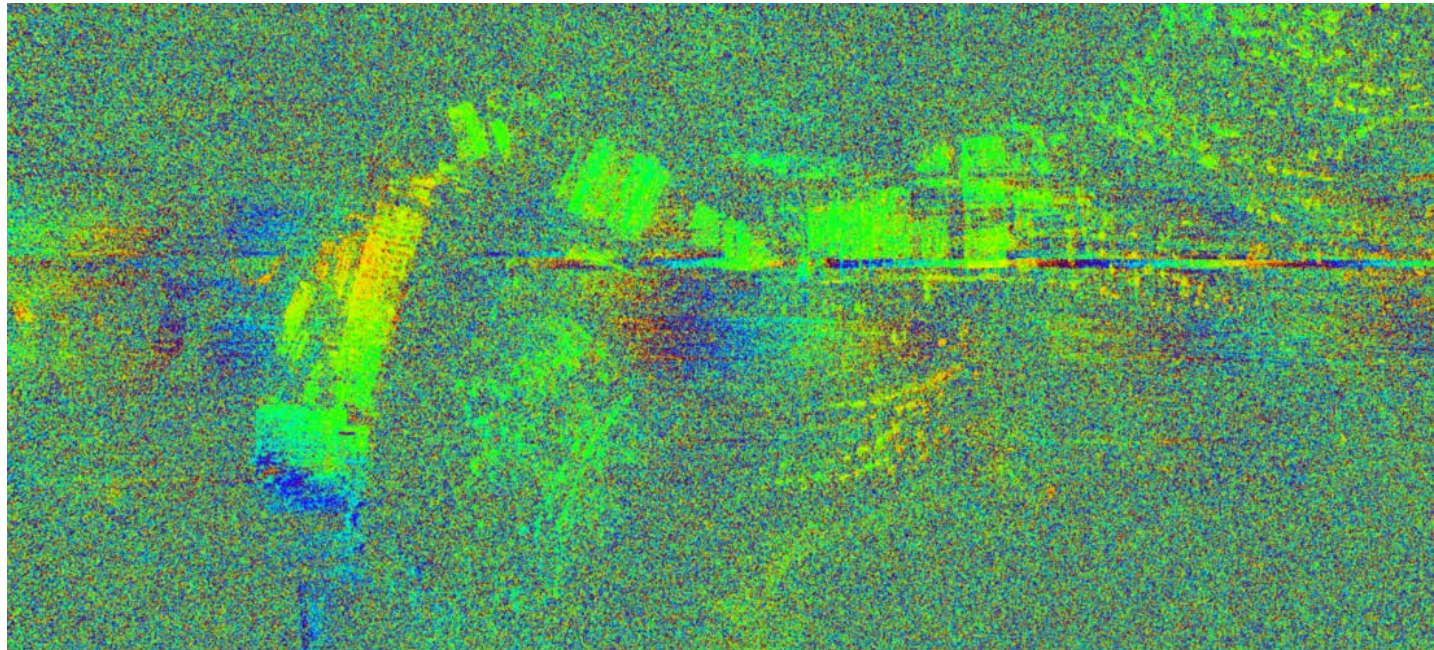


Overview of the project

Repeat-Pass Interferometry
Experiment

HITCHHIKER

First Spaceborne-Stationary
Experiment with TerraSAR-X

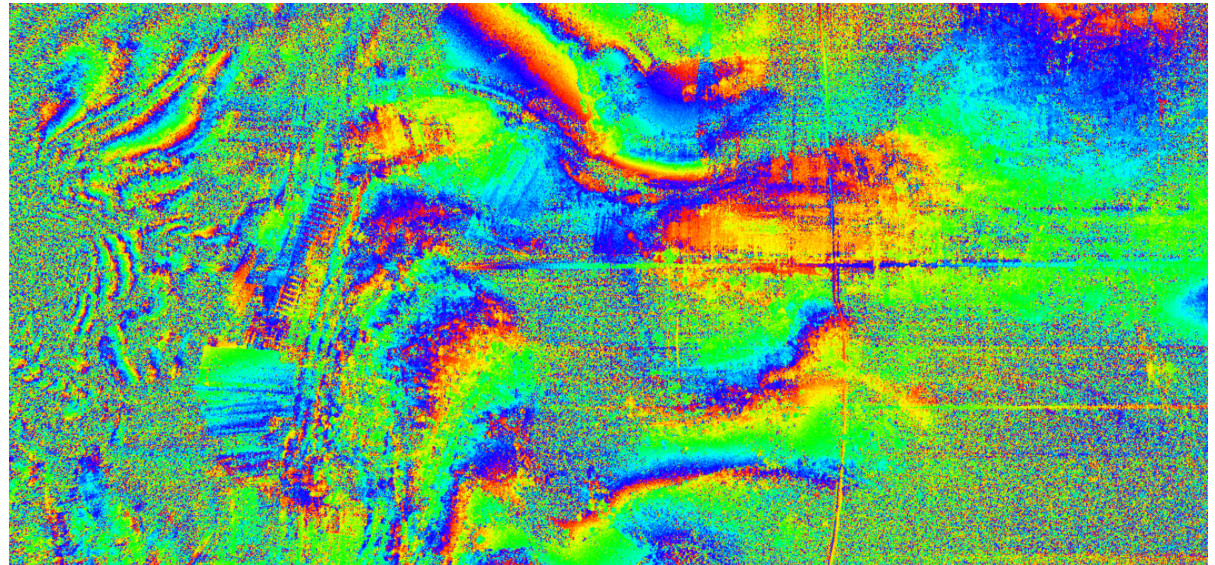


HITCHHIKER 4 Channel Extension



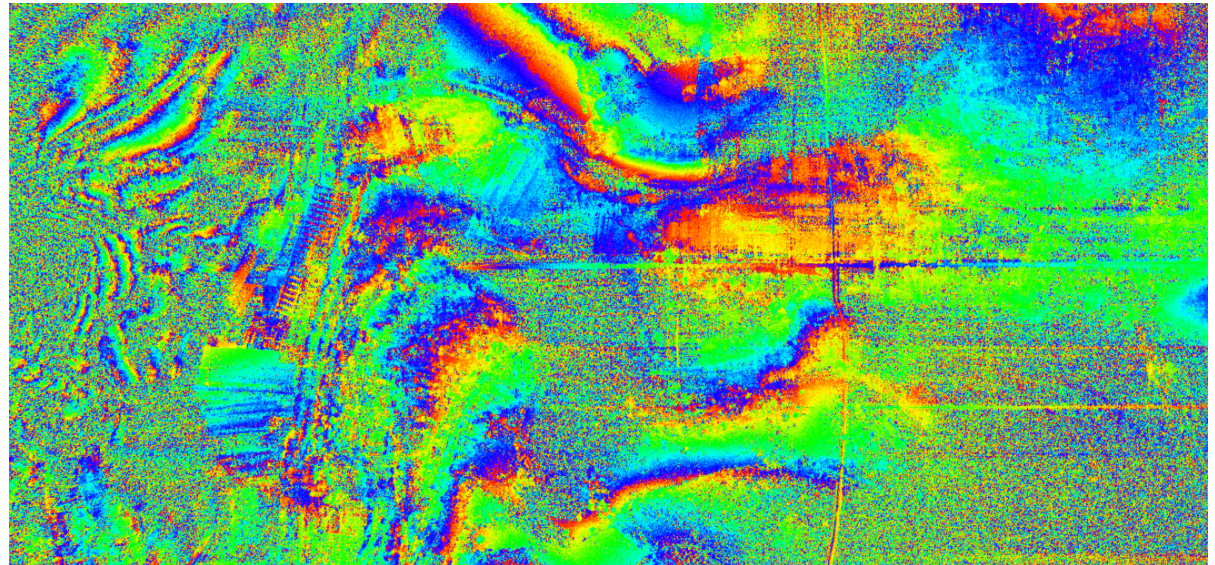
HITCHHIKER 4 Channel Extension

Single Pass Interferometry Experiment



HITCHHIKER
4 Channel Extension

Single Pass Interferometry
Experiment

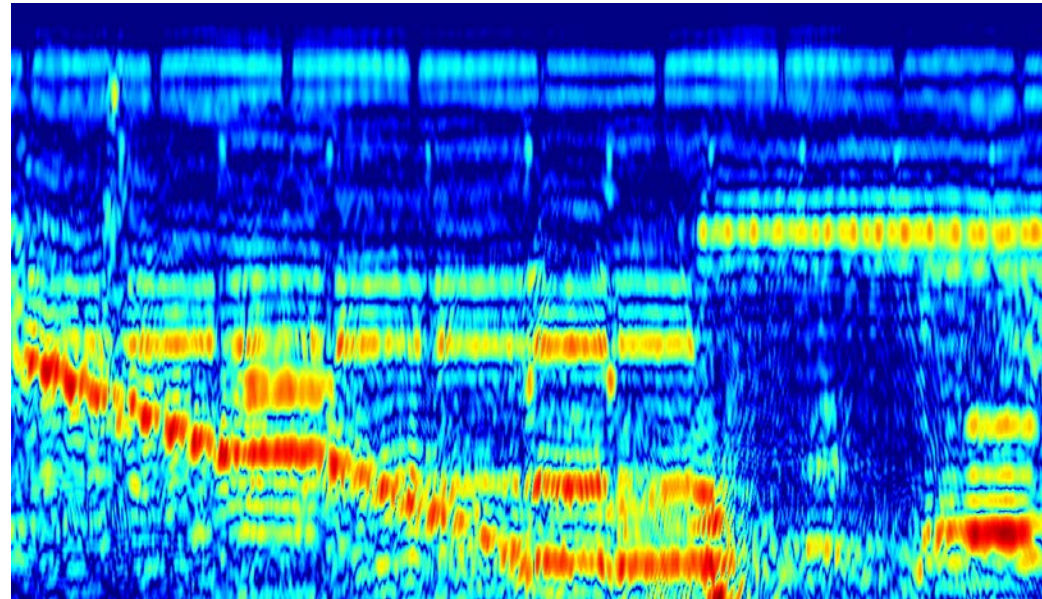


HITCHHIKER
Transmitter



HITCHHIKER
Transmitter

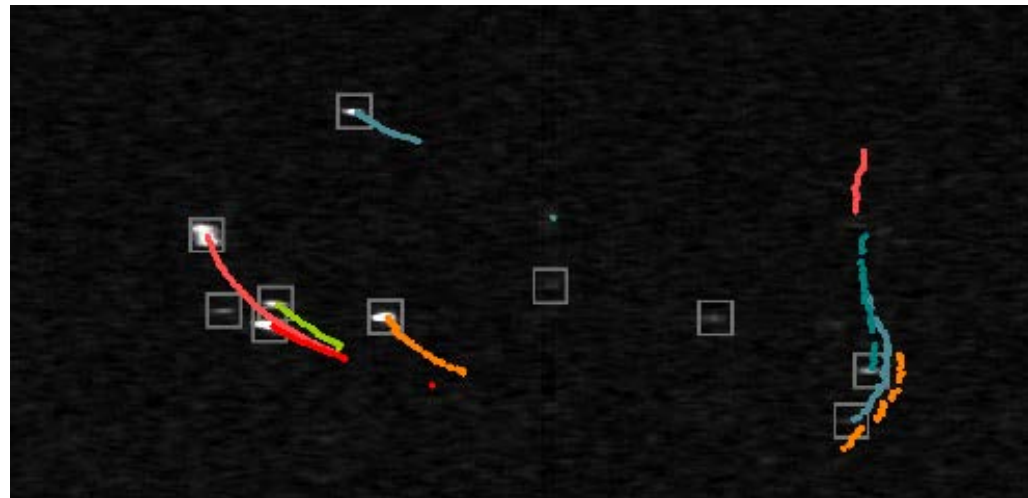
Indoor Noise SAR
Experiment



HITCHHIKER
Transmitter

Noise ISAR
Experiment

Indoor Noise SAR
Experiment

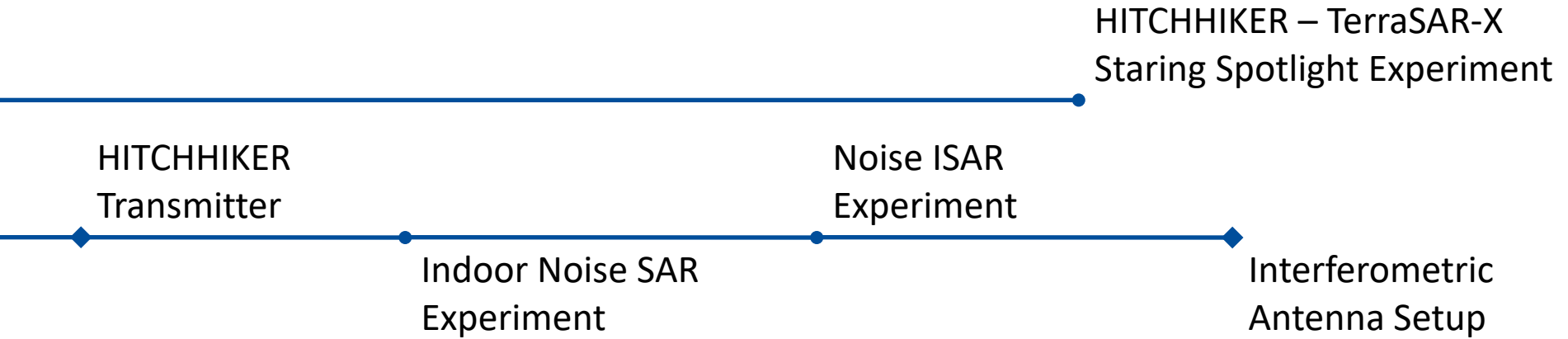


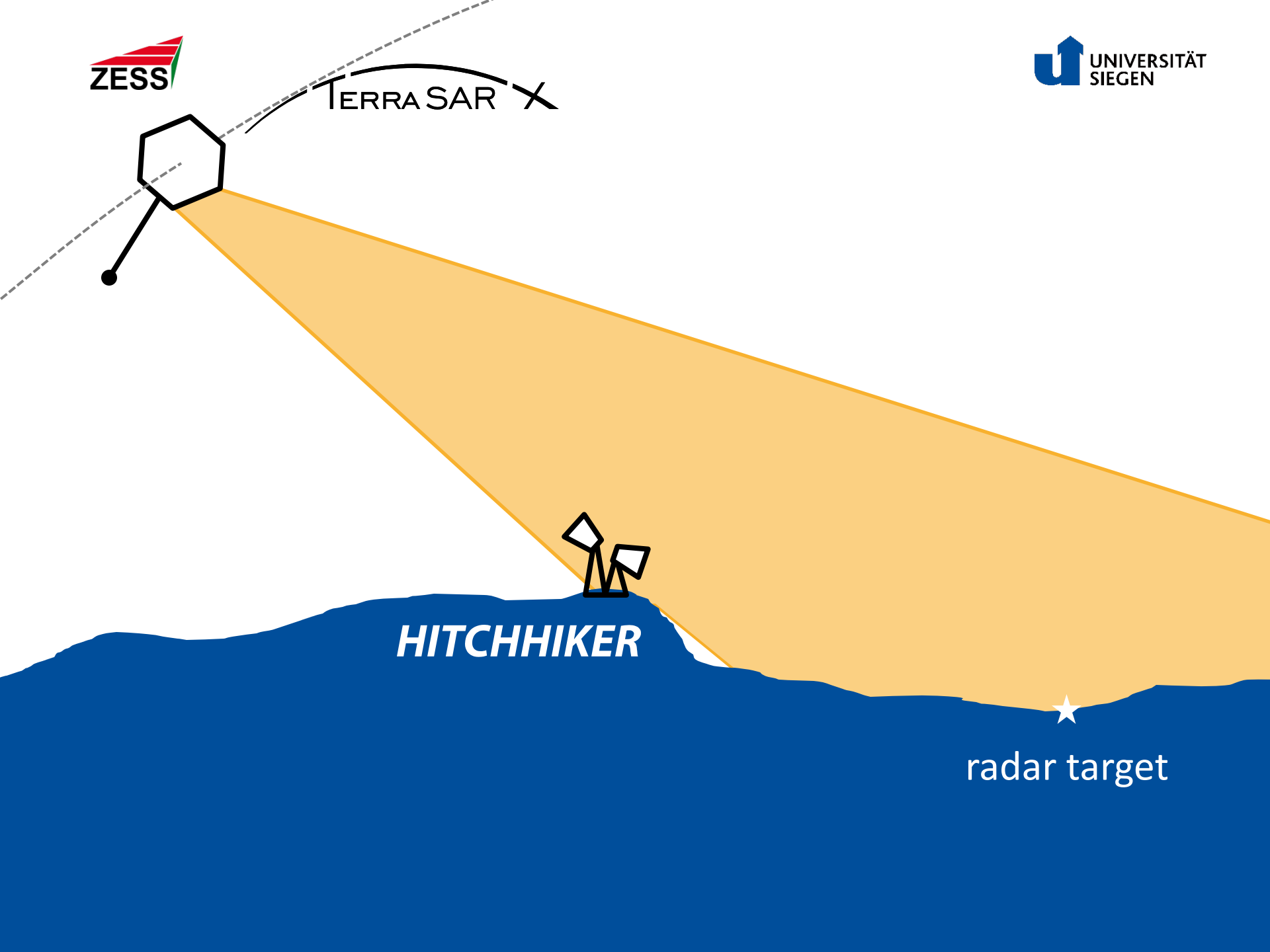
HITCHHIKER
Transmitter

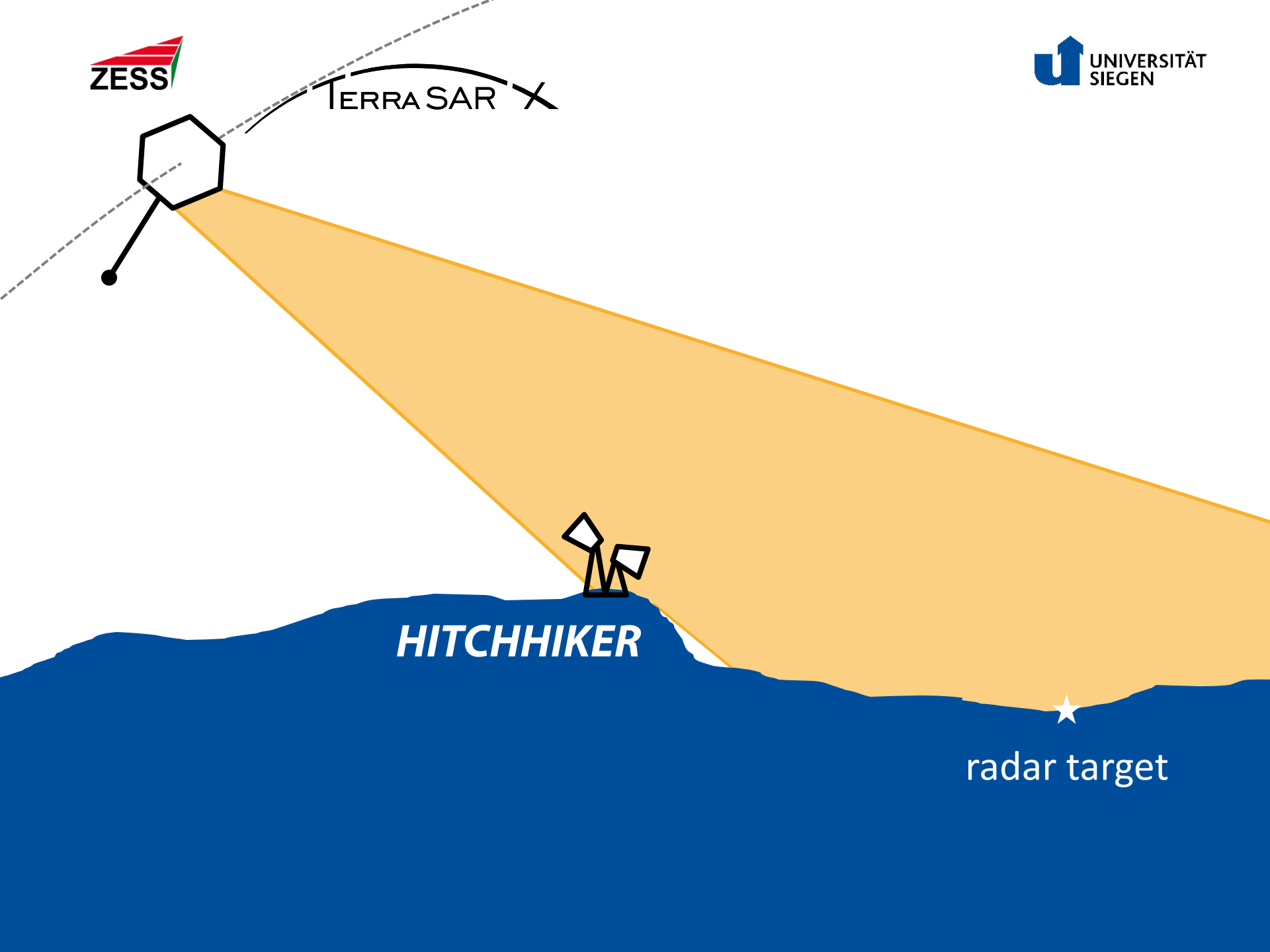
Indoor Noise SAR
Experiment

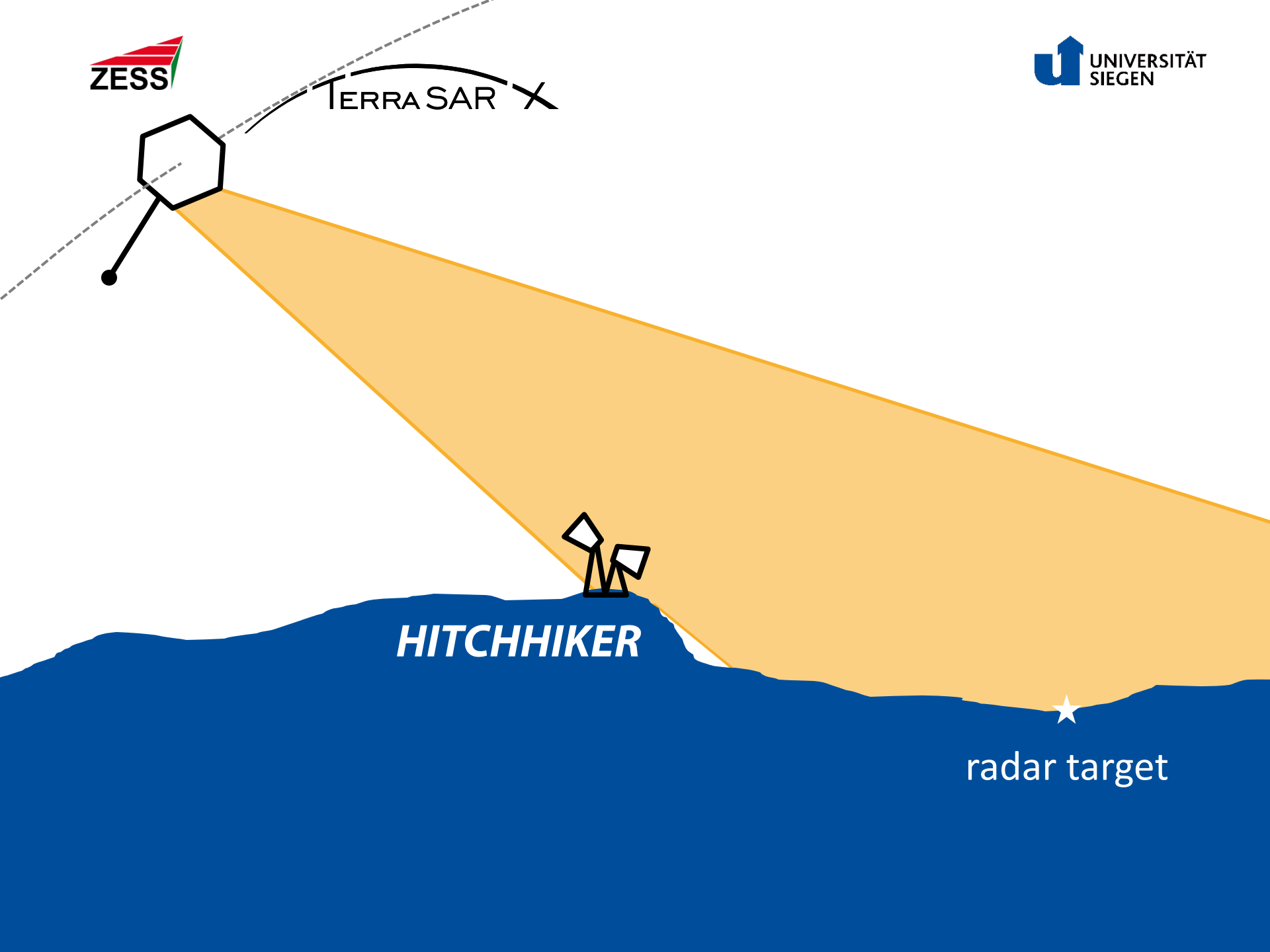
Noise ISAR
Experiment

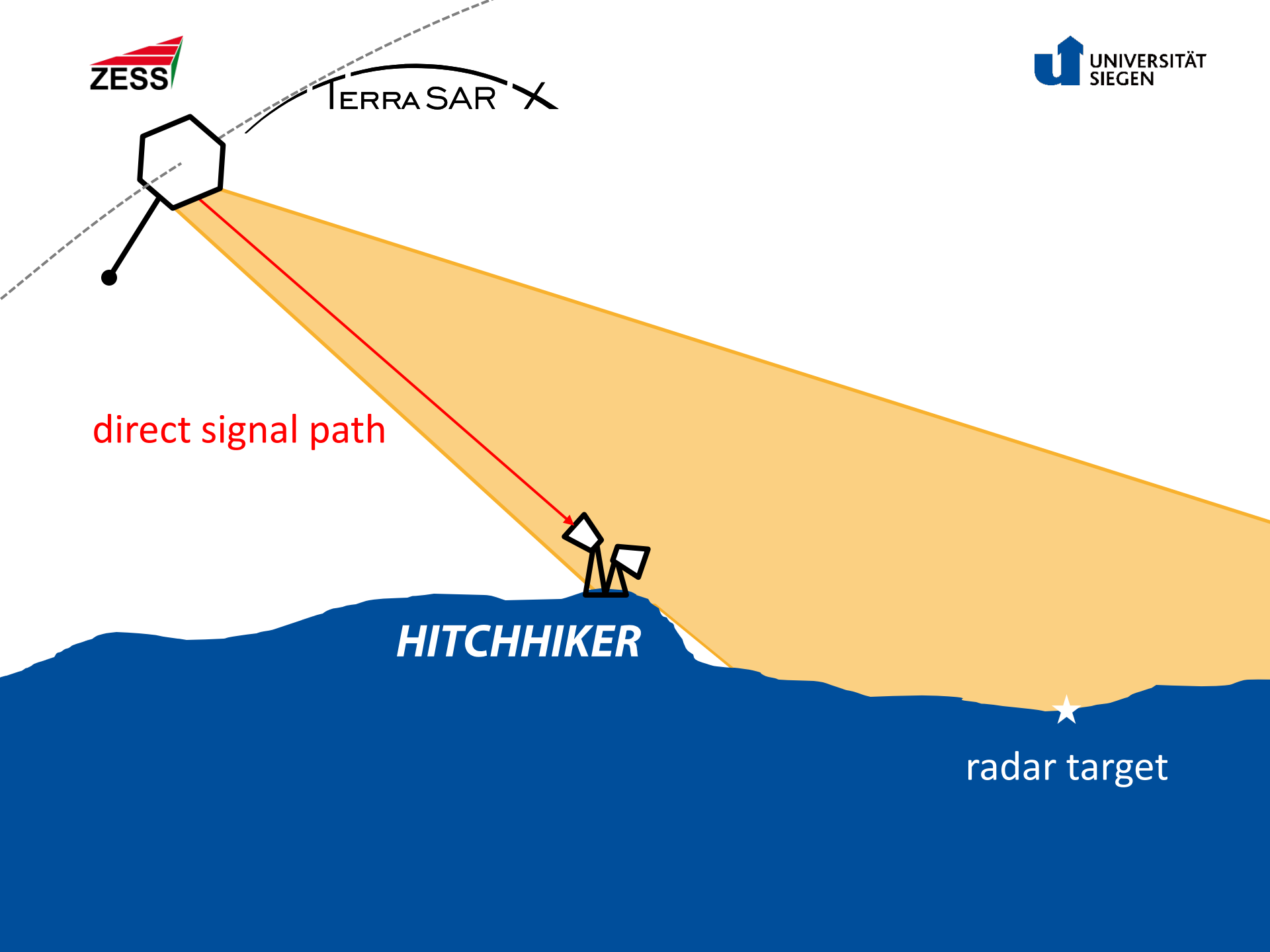
Interferometric
Antenna Setup

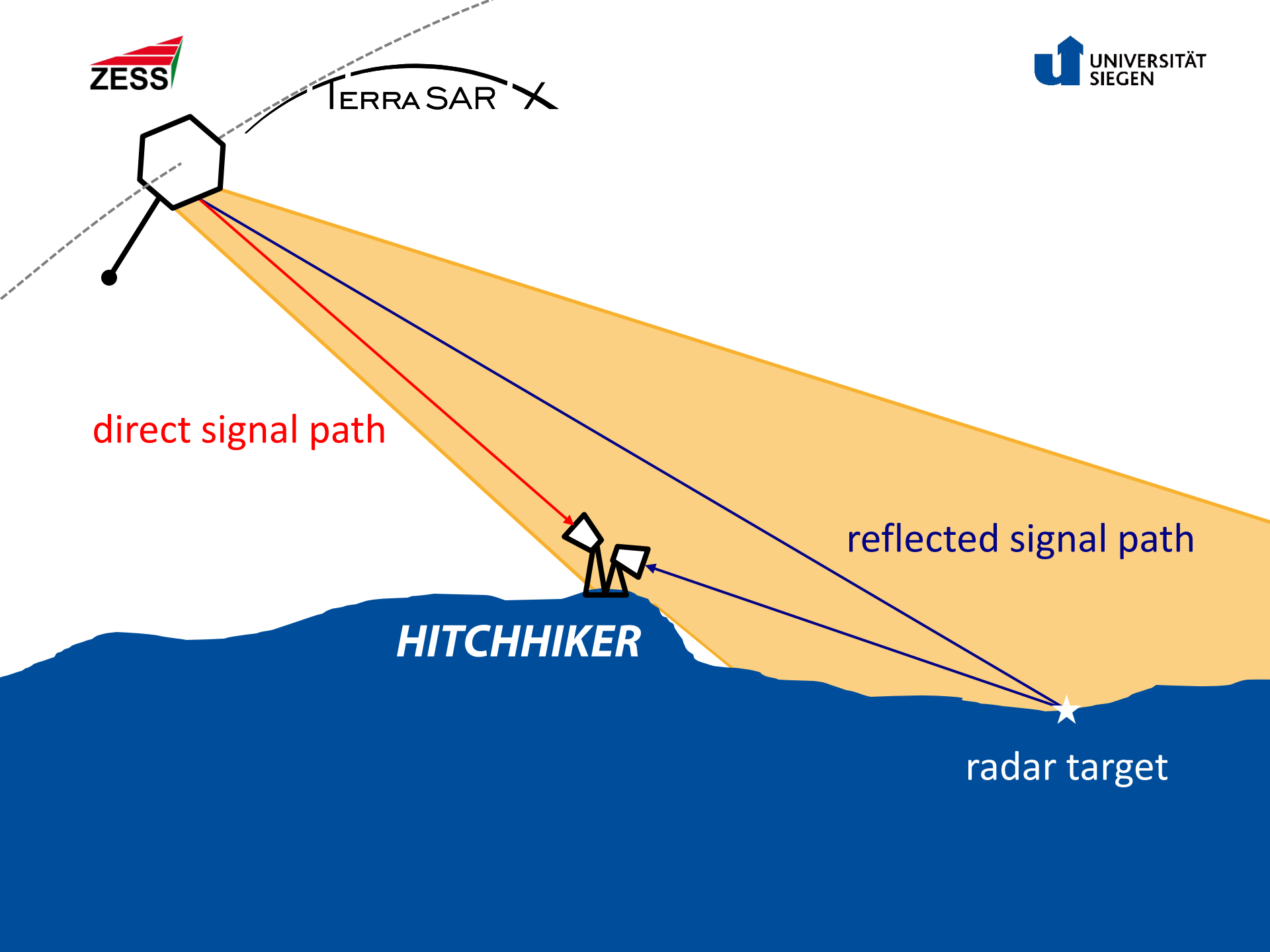


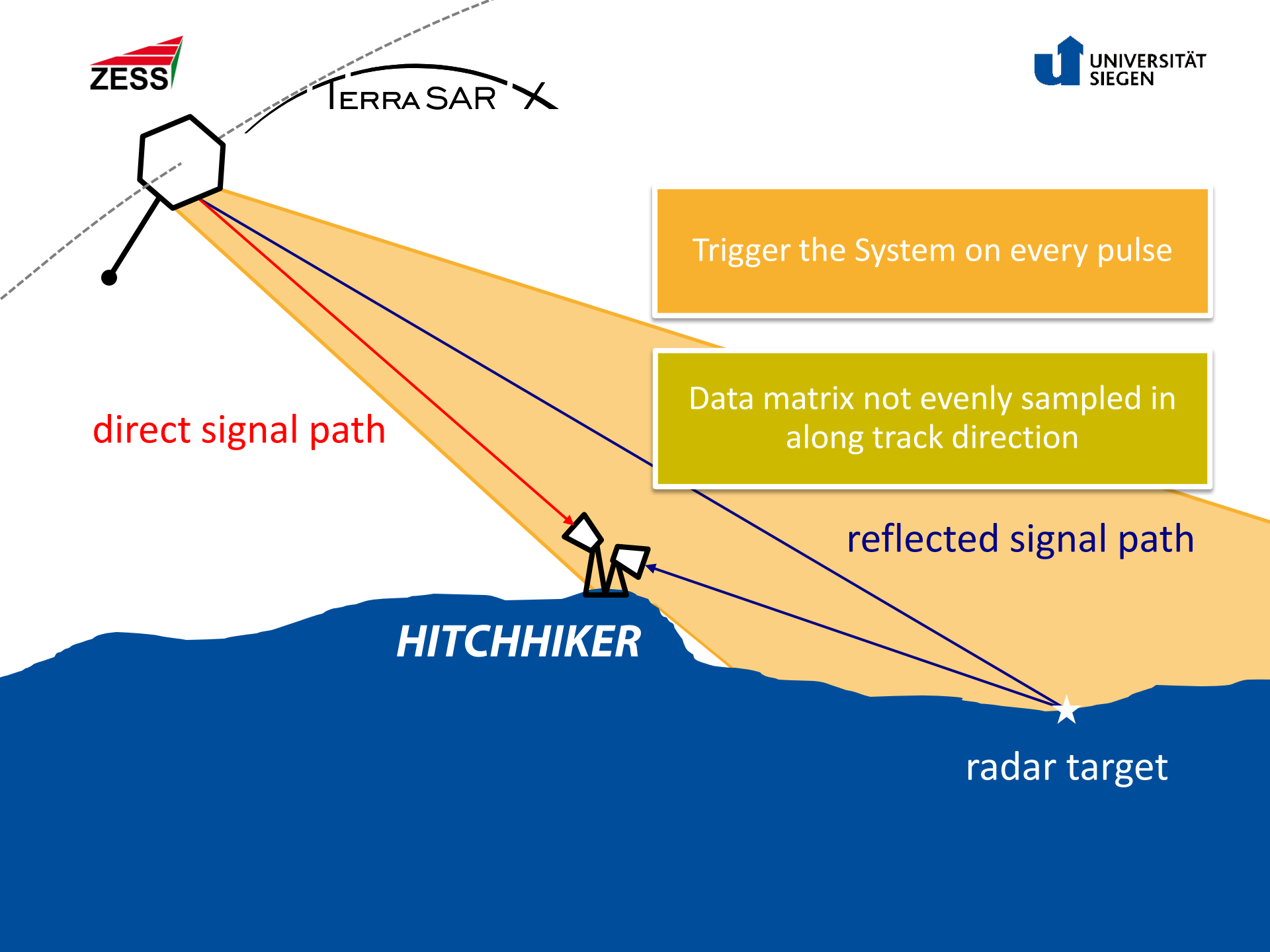


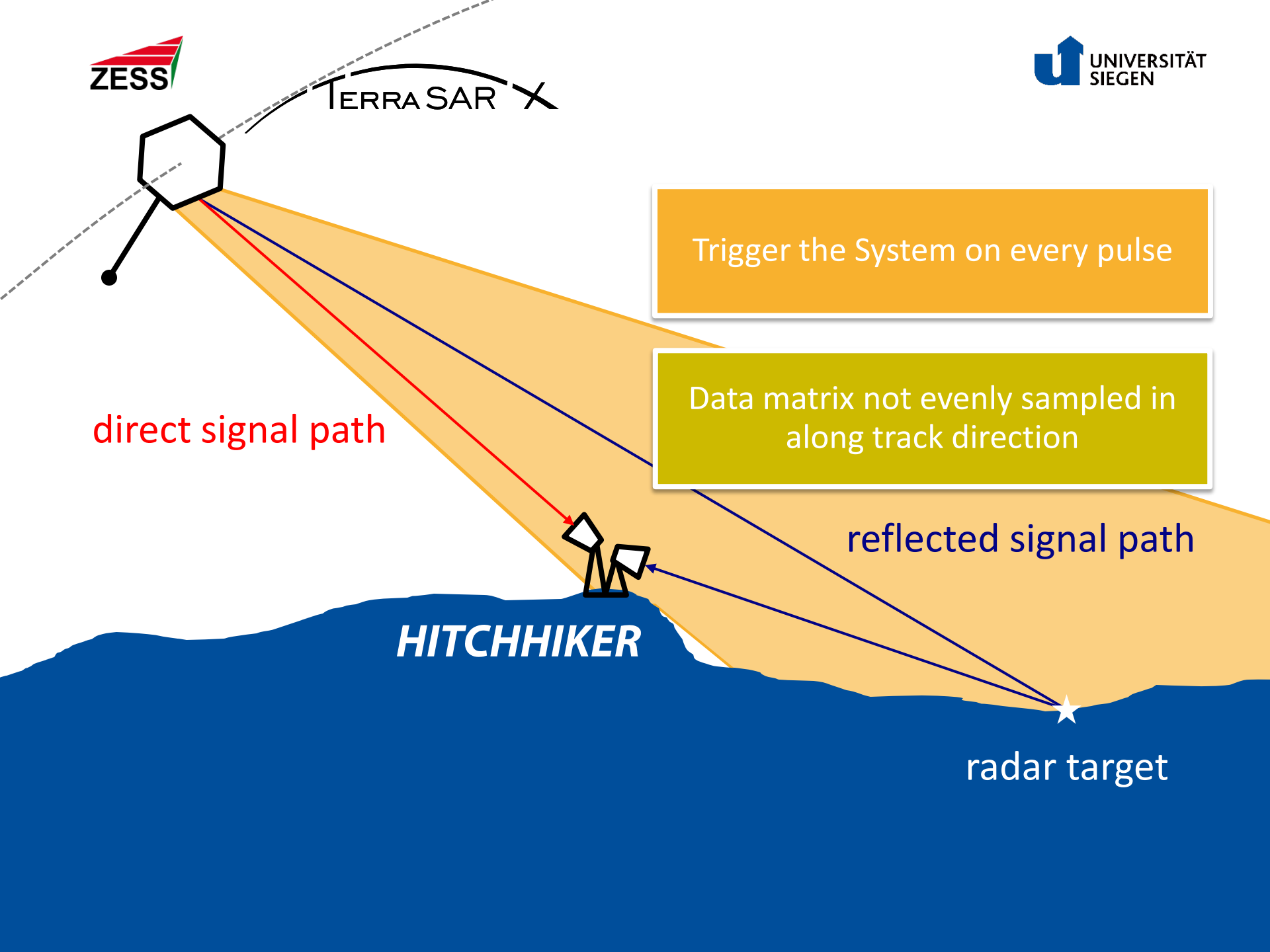




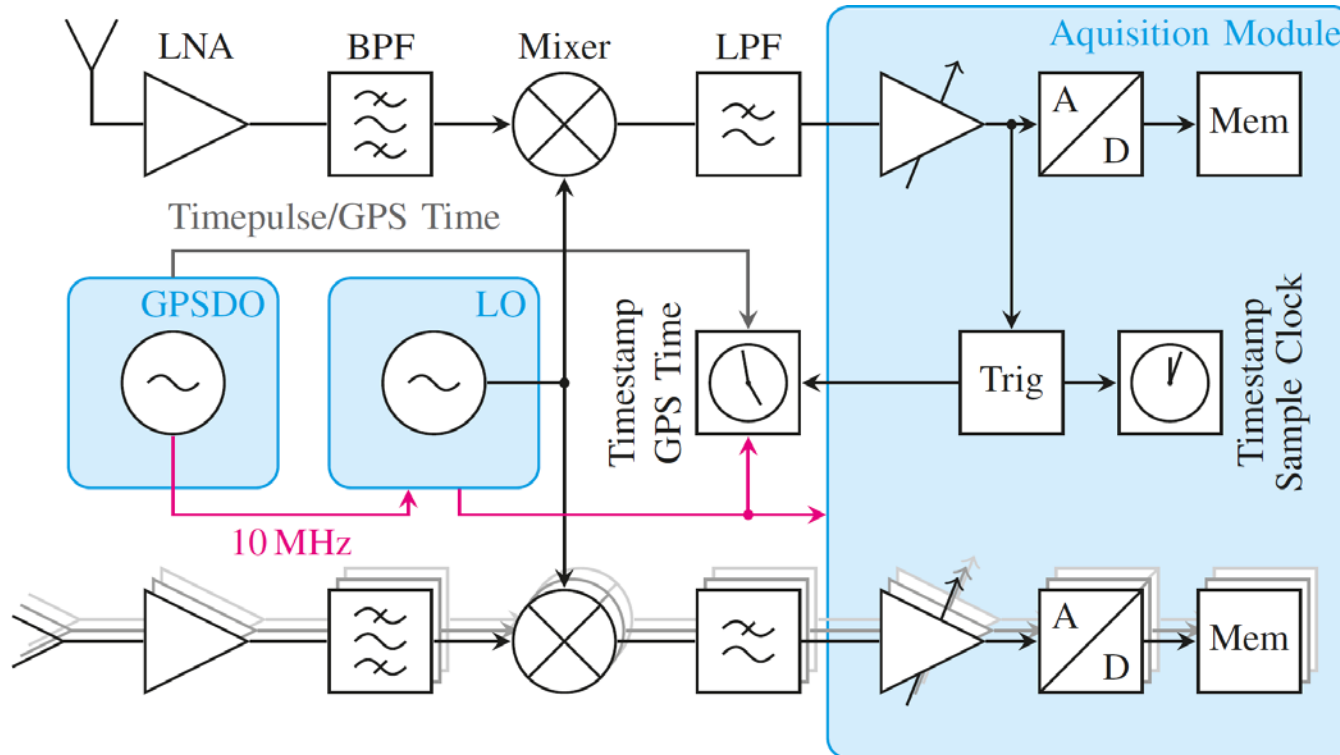






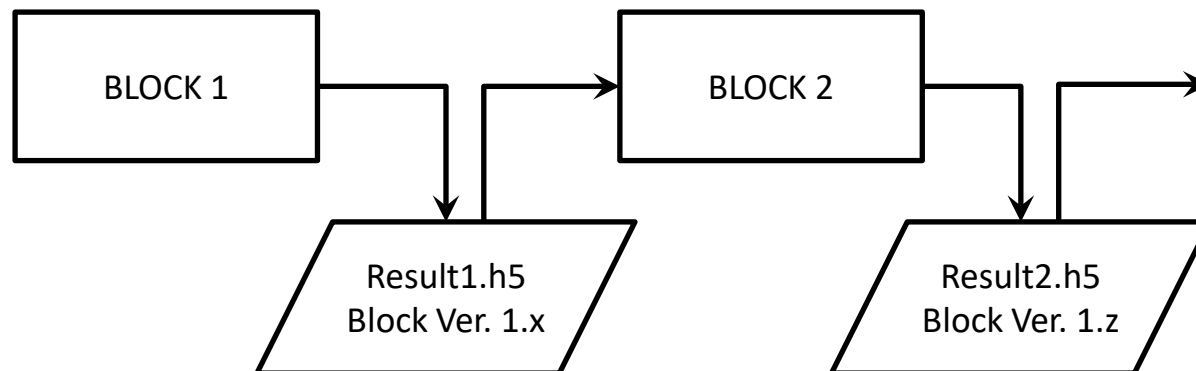


Receiver hardware synchronization

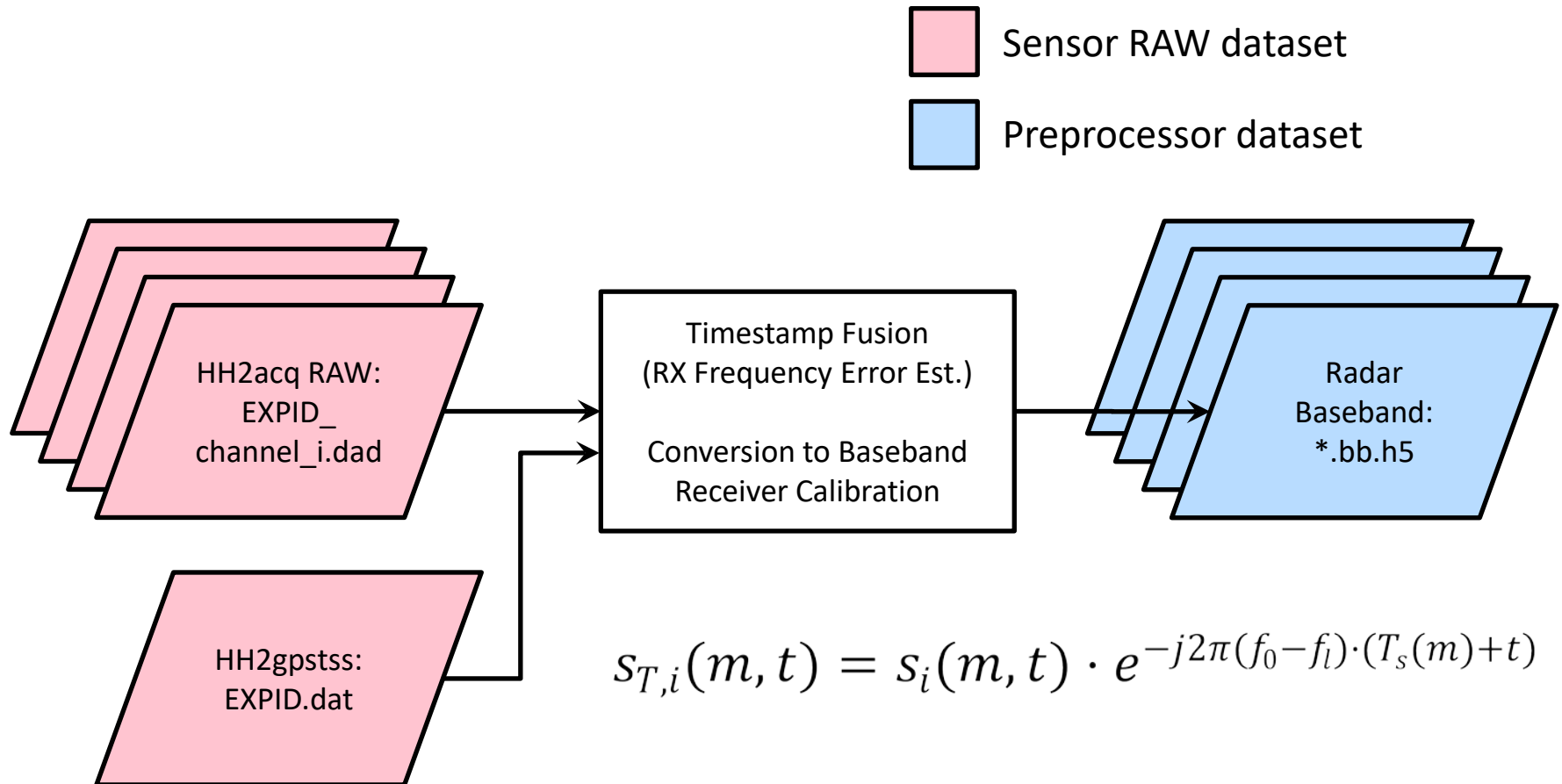


Processing framework

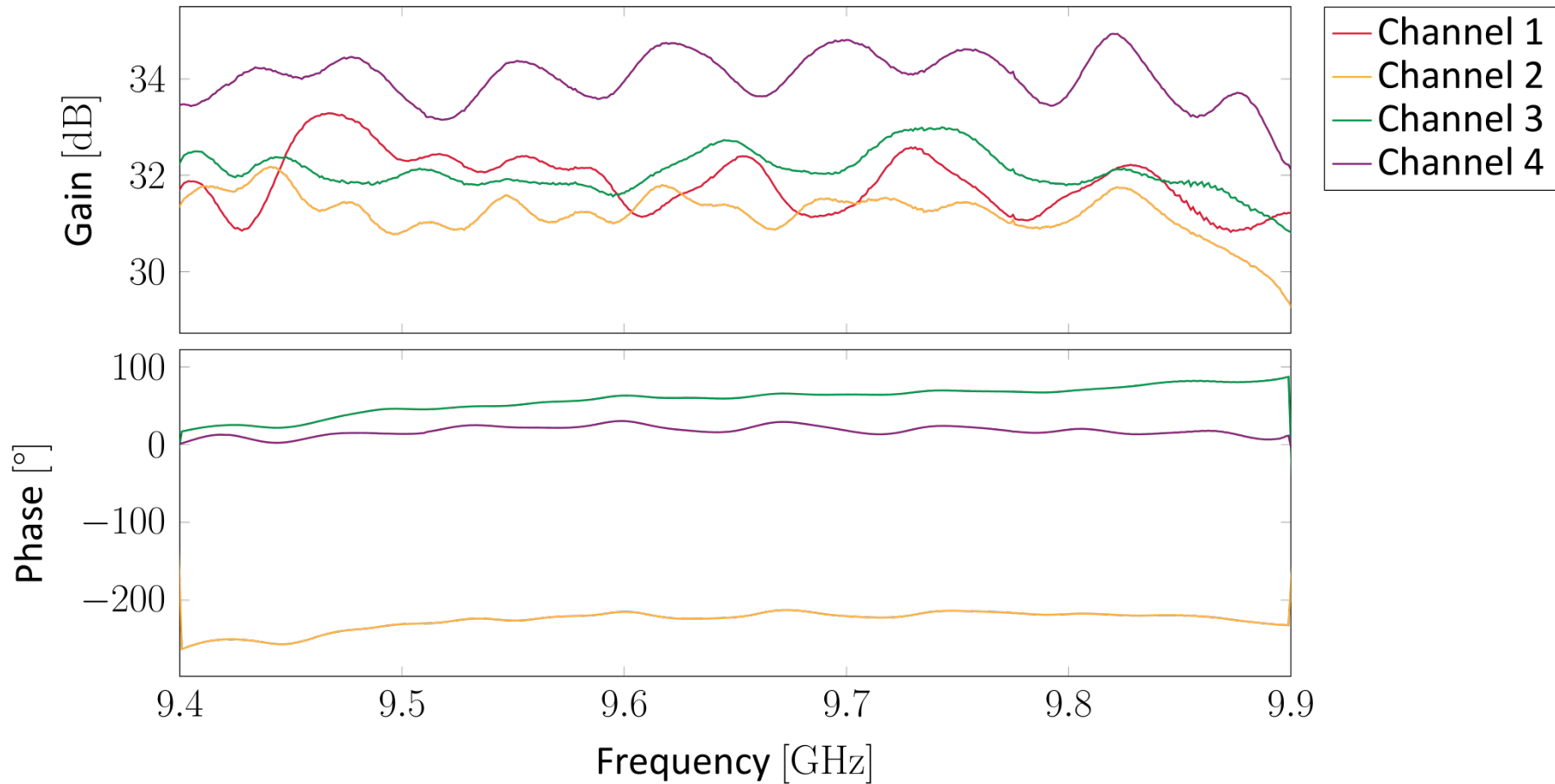
- Processing block implemented as Matlab class
 - parameters, input, output
 - Versioning and recursive validity check of data
- Use HDF 5 files as interface between blocks



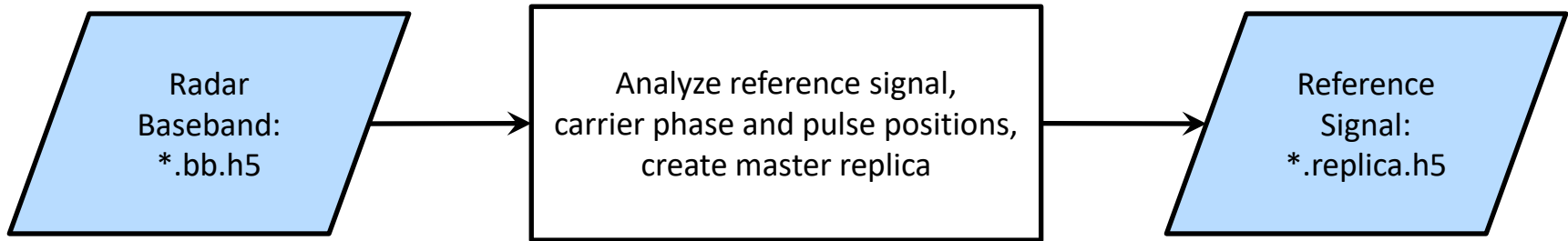
Baseband conversion



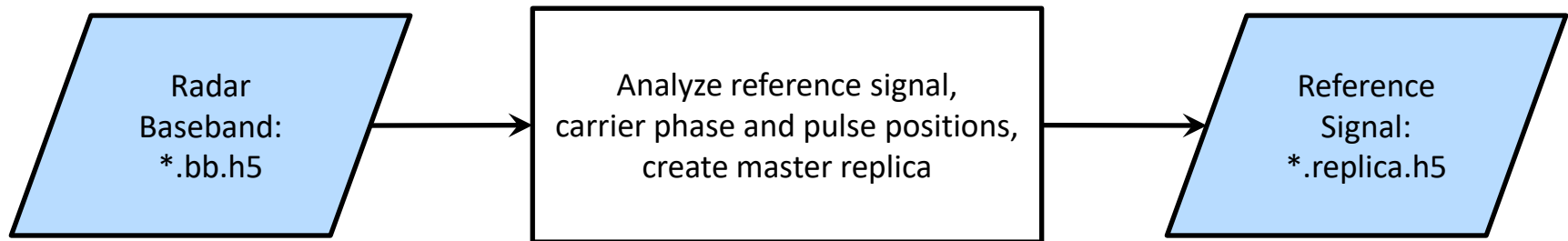
Baseband conversion and channel equalization



Analysis of the reference signal



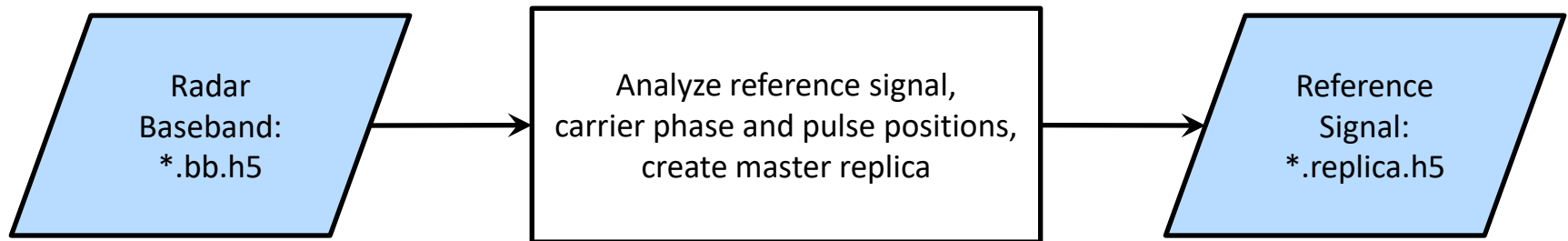
Analysis of the reference signal



$$s_d(m, t) = s_T(t - (\tau(m) - T_d(m))) \cdot e^{j2\pi(f'_0 - f_0) \cdot (t + T_s(m))} \cdot e^{-j2\pi f'_0 \tau(m)}$$

$$T_d(m) = T_s(m) - (T_0 + Z(m) \cdot T_i)$$

Analysis of the reference signal



$$s_d(m, t) = s_T(t - (\tau(m) - T_d(m))) \cdot e^{j2\pi(f'_0 - f_0) \cdot (t + T_s(m))} \cdot e^{-j2\pi f'_0 \tau(m)}$$

$$T_d(m) = T_s(m) - (T_0 + Z(m) \cdot T_i)$$

$$S_d(m, f) = (S_T(f) \cdot e^{-j2\pi f(\tau(m) - T_d(m))} \cdot e^{j2\pi(f'_0 - f_0) \cdot T_s(m)} \cdot e^{-j2\pi f'_0 \tau(m)}) * \delta(f - (f'_0 - f_0))$$

Analysis of the reference signal

$$\begin{aligned}\angle S_d(m, f) \cdot S_d^*(M, f) = & 2\pi f(T_d(m) - T_d(M)) - 2\pi f(\tau(m) - \tau(M)) \\ & - 2\pi f_0(\tau(m) - \tau(M)) \\ & + 2\pi(f'_0 - f_0)(Z(m) - Z(M)) \cdot T_i\end{aligned}$$

Analysis of the reference signal

$$\begin{aligned}\angle S_d(m, f) \cdot S_d^*(M, f) &= 2\pi f(T_d(m) - T_d(M)) - 2\pi f(\tau(m) - \tau(M)) \\ &\quad - 2\pi f_0(\tau(m) - \tau(M)) \\ &\quad + 2\pi(f'_0 - f_0)(Z(m) - Z(M)) \cdot T_i\end{aligned}$$

$$\begin{aligned}\Delta T(m) &= -\frac{1}{2\pi} \frac{d}{df} \angle S_d(m, f) \cdot S_d^*(M, f) \\ &= \tau(m) - \tau(M) \\ &\quad - (T_d(m) - T_d(M))\end{aligned}$$

Analysis of the reference signal

$$\begin{aligned}\angle S_d(m, f) \cdot S_d^*(M, f) &= 2\pi f(T_d(m) - T_d(M)) - 2\pi f(\tau(m) - \tau(M)) \\ &\quad - 2\pi f_0(\tau(m) - \tau(M)) \\ &\quad + 2\pi(f'_0 - f_0)(Z(m) - Z(M)) \cdot T_i\end{aligned}$$

$$\begin{aligned}\Delta T(m) &= -\frac{1}{2\pi} \frac{d}{df} \angle S_d(m, f) \cdot S_d^*(M, f) \\ &= \tau(m) - \tau(M) \\ &\quad - (T_d(m) - T_d(M))\end{aligned}$$

$$\begin{aligned}\varphi_T(m) &= \angle S_d(m, f) \cdot S_d^*(M, f)|_{f=0} \\ &= 2\pi(f'_0 - f_0)(Z(m) - Z(M)) \cdot T_i \\ &\quad - 2\pi f_0(\tau(m) - \tau(M))\end{aligned}$$

Calculation of the replica

$$\begin{aligned}
 S'_T(f) &= \sum_m S_d(m, f) \cdot e^{j2\pi f \Delta T(m)} e^{-j\varphi_T(m)} \\
 &= C \cdot S_T(f - (f'_0 - f_0)) e^{-j2\pi f (\tau(M) - T_d(M))} \\
 &\quad \cdot e^{j2\pi (f'_0 - f_0) Z(M) \cdot T_i} e^{-j2\pi f_0 \tau(M)}
 \end{aligned}$$

Calculation of the replica

$$\begin{aligned}
 S'_T(f) &= \sum_m S_d(m, f) \cdot e^{j2\pi f \Delta T(m)} e^{-j\varphi_T(m)} \\
 &= C \cdot S_T(f - (f'_0 - f_0)) e^{-j2\pi f (\tau(M) - T_d(M))} \\
 &\quad \cdot e^{j2\pi (f'_0 - f_0) Z(M) \cdot T_i} e^{-j2\pi f_0 \tau(M)}
 \end{aligned}$$

Correct relative **delay** and **phase** to reference pulse M

Integrate over all completely received pulses

Calculation of the replica

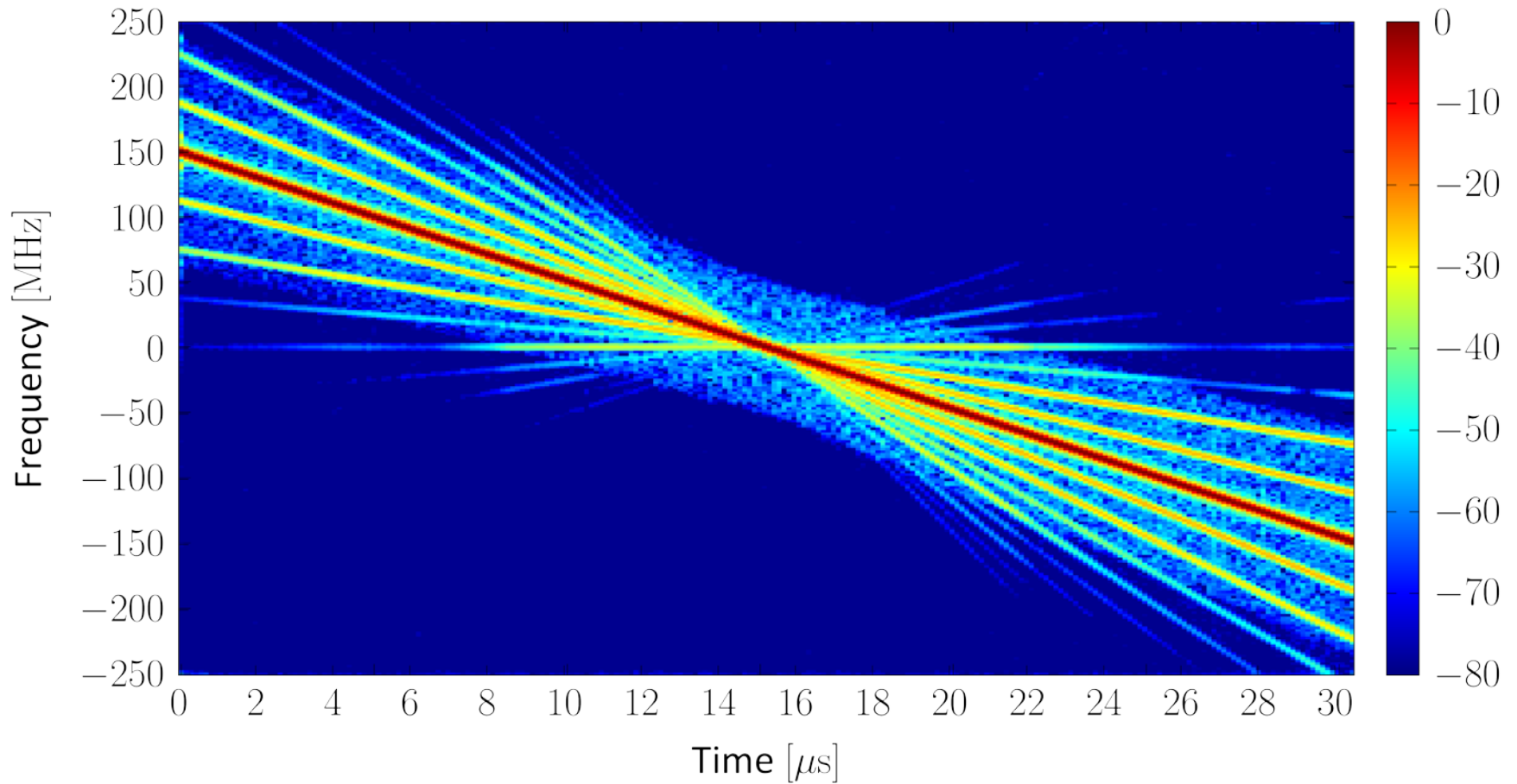
$$\begin{aligned}
 S'_T(f) &= \sum_m S_d(m, f) \cdot e^{j2\pi f \Delta T(m)} e^{-j\varphi_T(m)} \\
 &= C \cdot S_T(f - (f'_0 - f_0)) e^{-j2\pi f (\tau(M) - T_d(M))} \\
 &\quad \cdot e^{j2\pi (f'_0 - f_0) Z(M) \cdot T_i} e^{-j2\pi f_0 \tau(M)}
 \end{aligned}$$

Correct relative **delay** and **phase** to reference pulse M

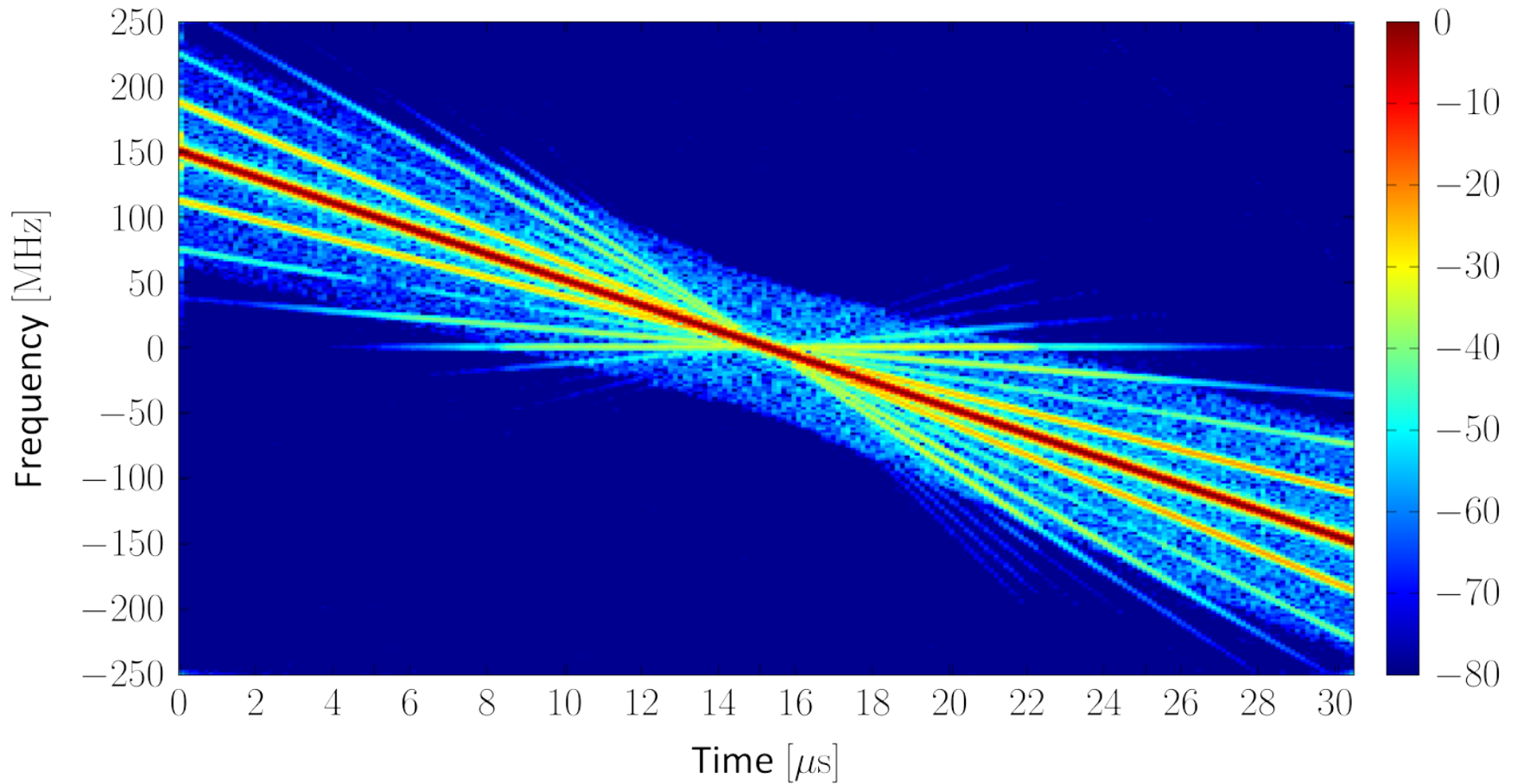
Integrate over all completely received pulses

$$T_r(m) = T_s(m) + \Delta T(m)$$

TerraSAR-X Signal



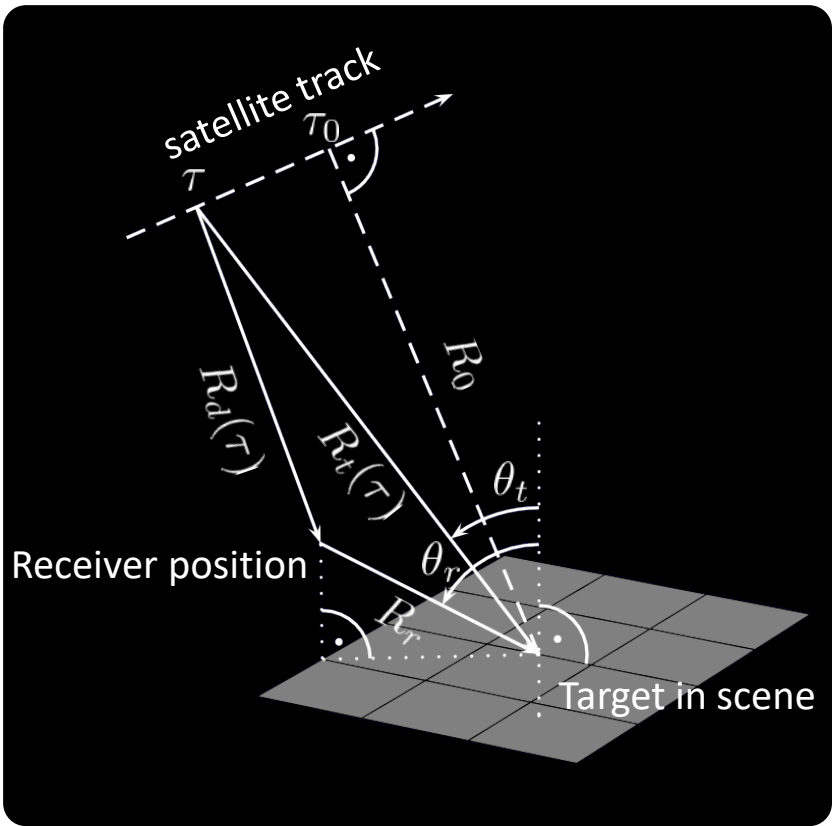
TanDEM-X Signal



Bistatic SAR Image Formation

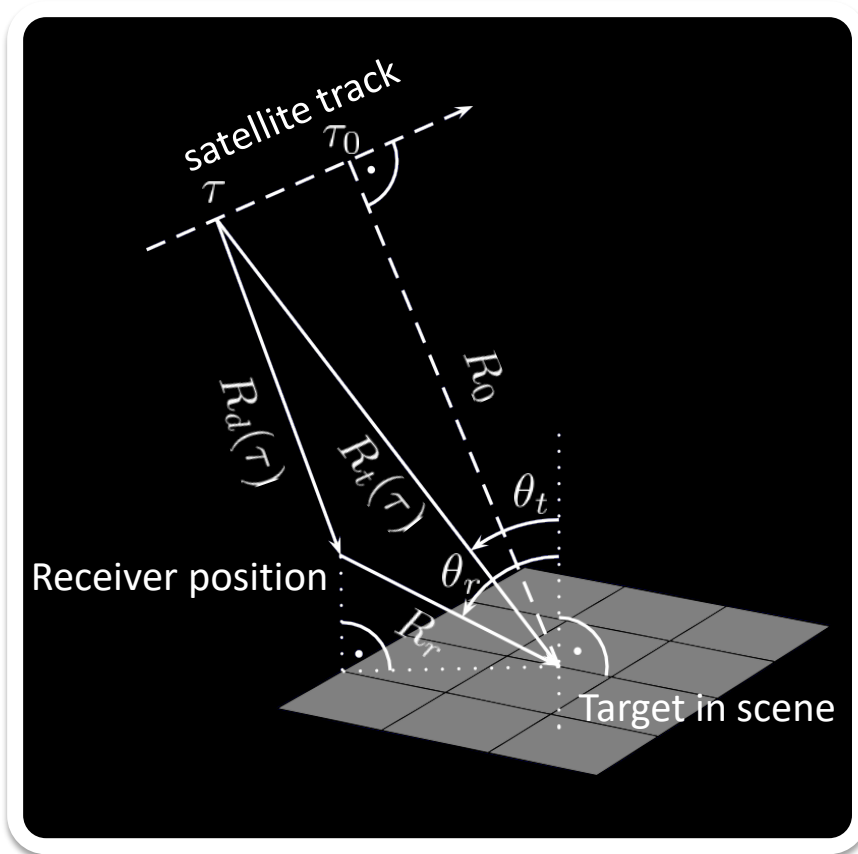
Bistatic Geometry

Resolution



Bistatic Geometry

Resolution

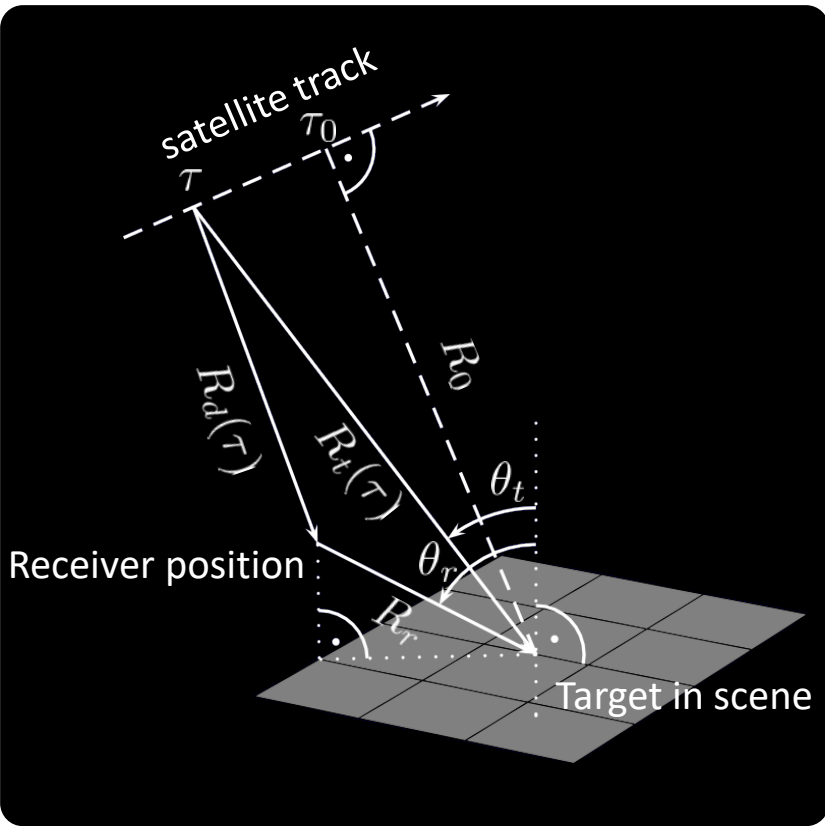


Across-Track Resolution

$$\delta_c = \frac{c}{B \cdot (\sin \theta_t + \sin \theta_r)}$$

Bistatic Geometry

Resolution



Across-Track Resolution

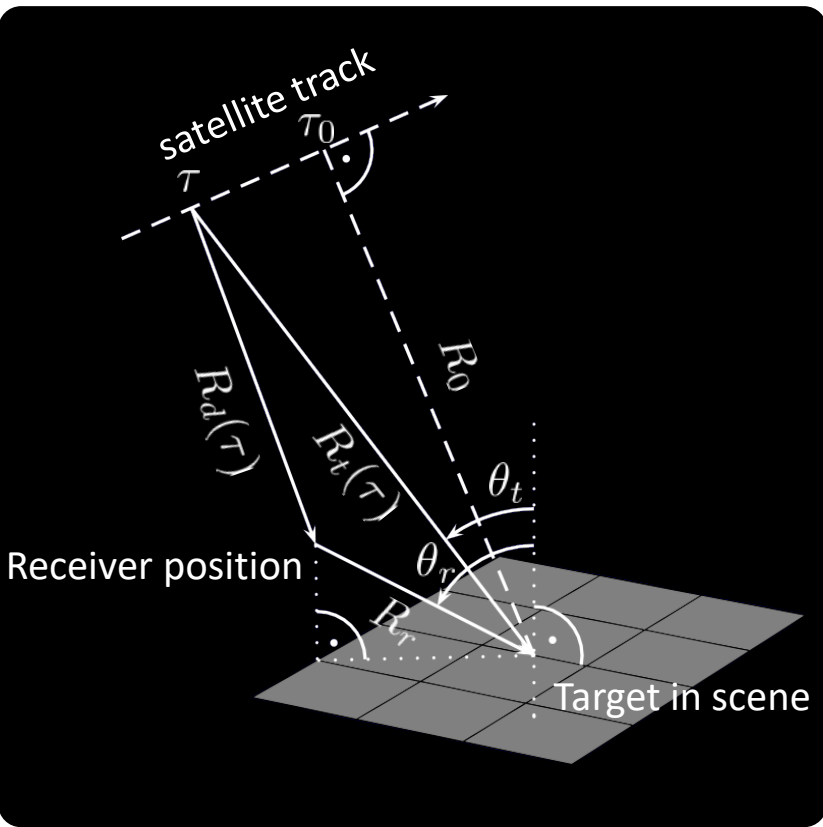
$$\delta_c = \frac{c}{B \cdot (\sin \theta_t + \sin \theta_r)}$$

Along-Track Resolution

$$\delta_a = \frac{\lambda_0}{2} \sqrt{\frac{4R_0^2}{(v \cdot T_i)^2} + 1}$$

Bistatic Geometry

Resolution



Across-Track Resolution

$$\delta_c = \frac{c}{B \cdot (\sin \theta_t + \sin \theta_r)}$$

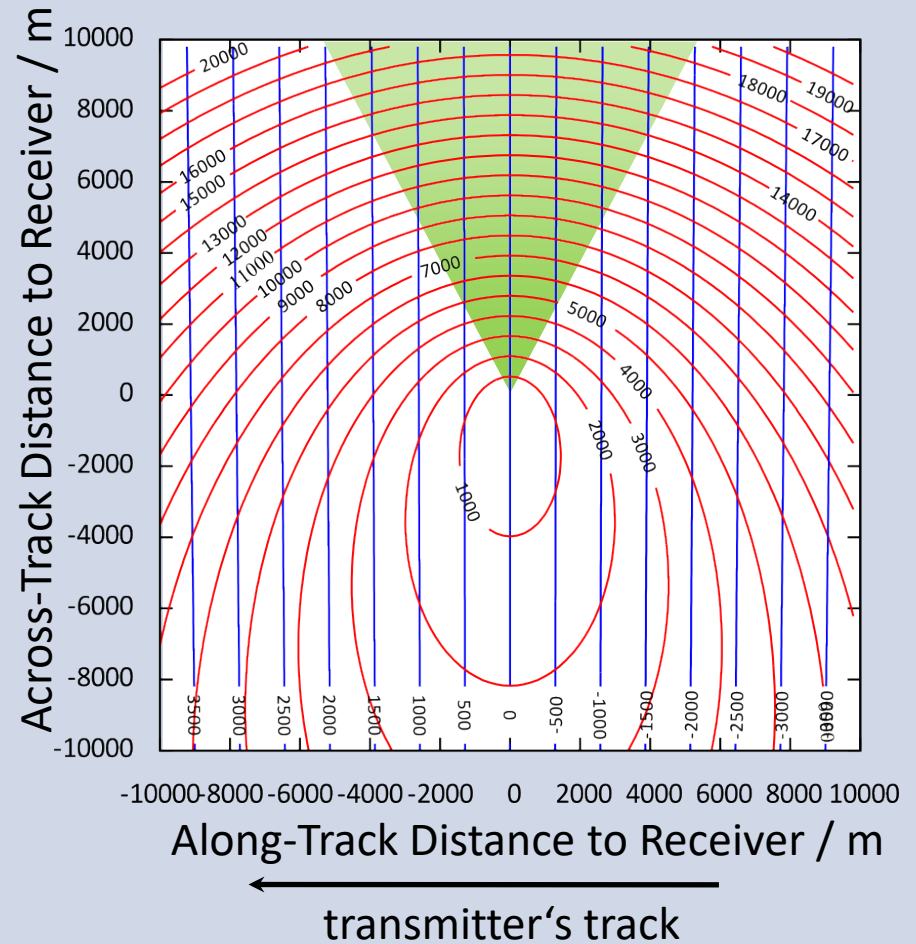
Along-Track Resolution

$$\delta_a = \frac{\lambda_0}{2} \sqrt{\frac{4R_0^2}{(v \cdot T_i)^2} + 1}$$

- λ_0 : wavelength
- R_0 : min. range target - transmitter
- v : transmitter's velocity
- T_i : illumination time

Bistatic Geometry

Iso-range and iso-doppler contours

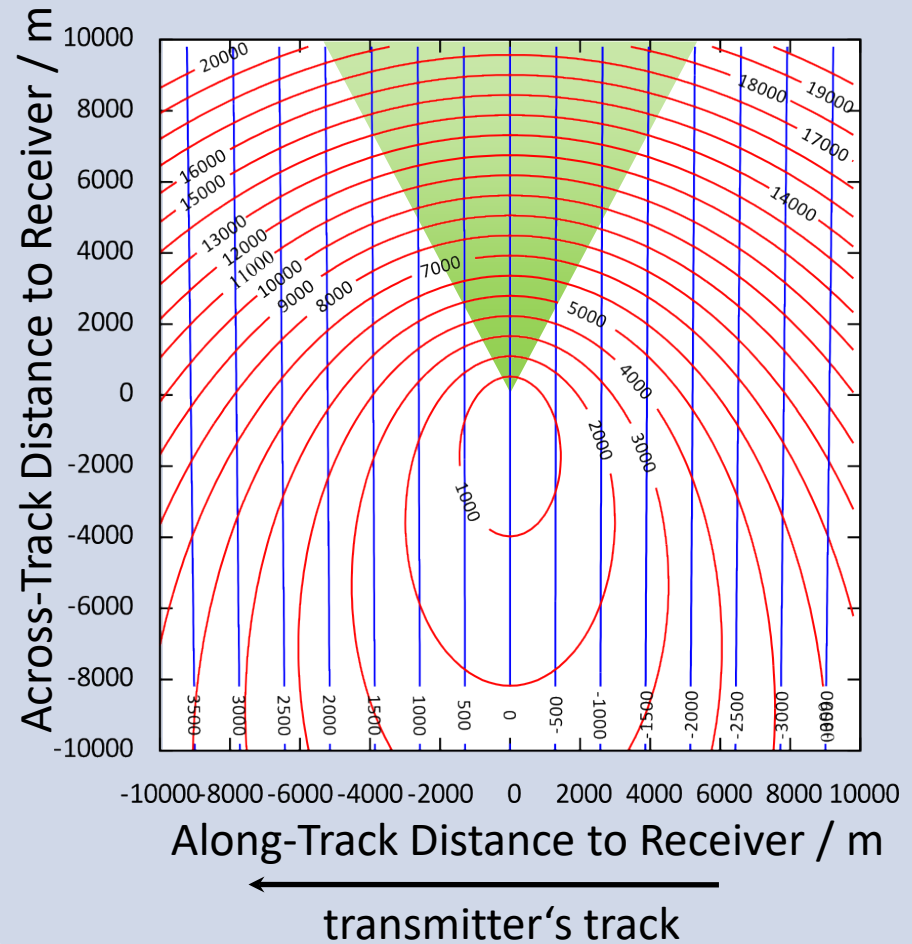


Bistatic Geometry

Iso-range and iso-doppler contours

Across-track resolution:

0.56 m



Bistatic Geometry

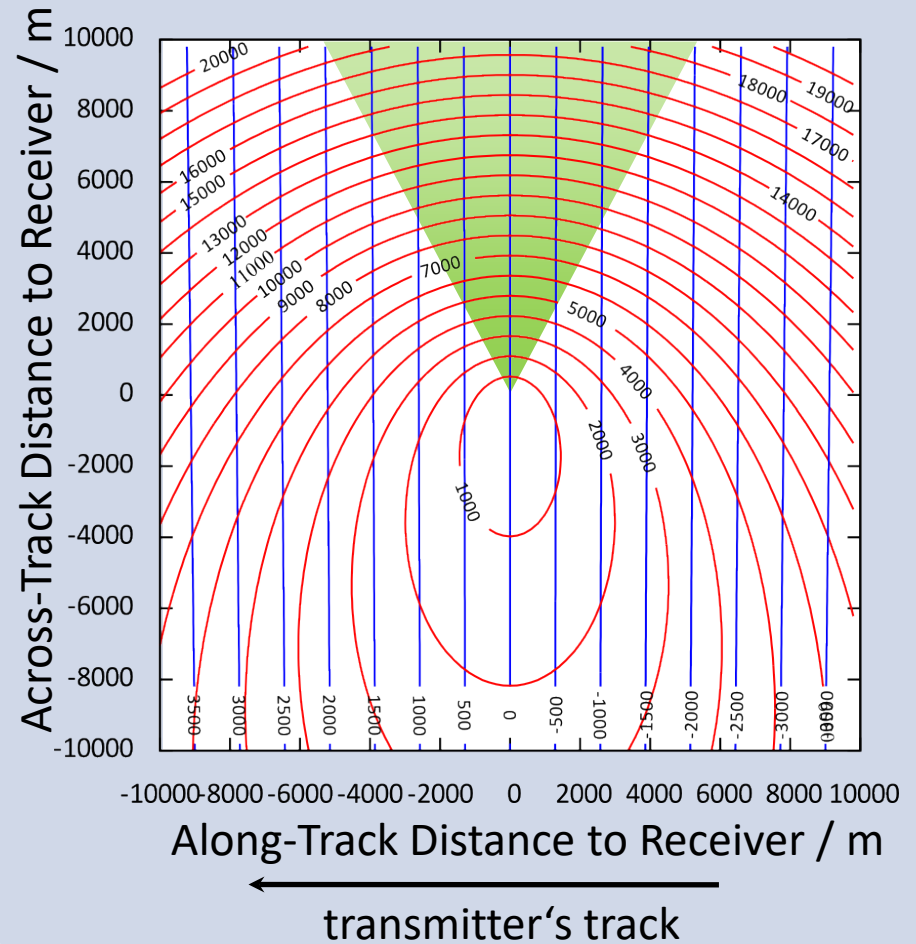
Iso-range and iso-doppler contours

Across-track resolution:

0.56 m

Along-track resolution
(HS mode):

1.40 m



Bistatic Geometry

Iso-range and iso-doppler contours

Across-track resolution:

0.56 m

Along-track resolution
(HS mode):

1.40 m

**Best resolution is achieved
when transmitter and receiver are
looking into the same direction**

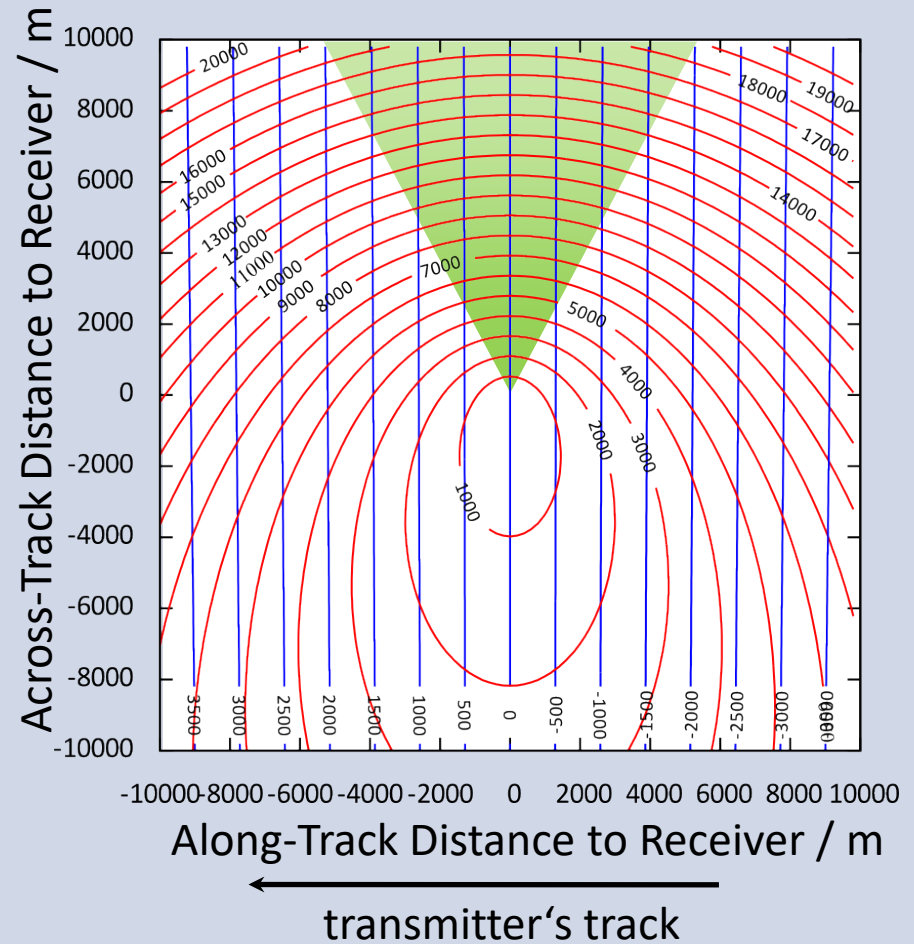
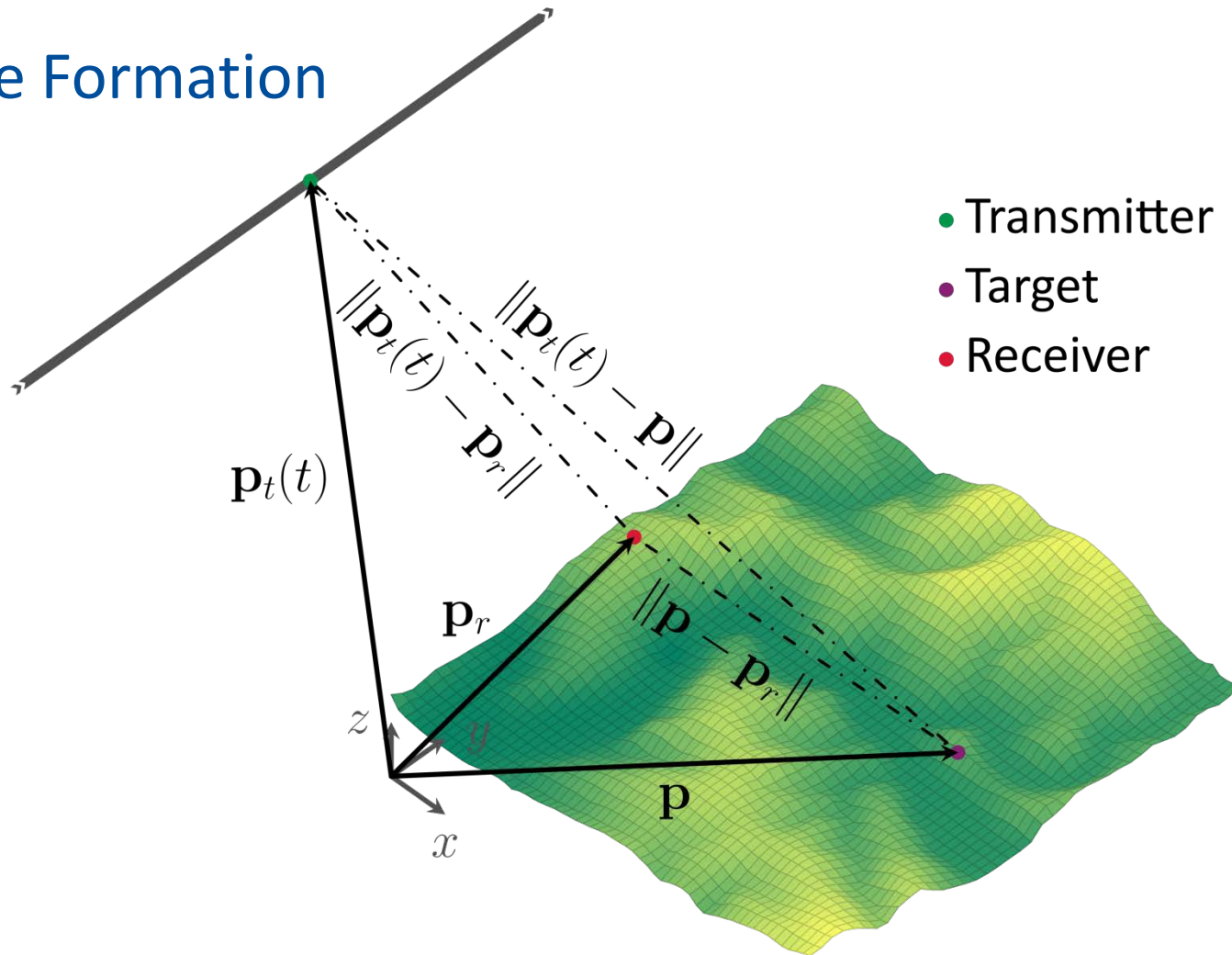


Image Formation

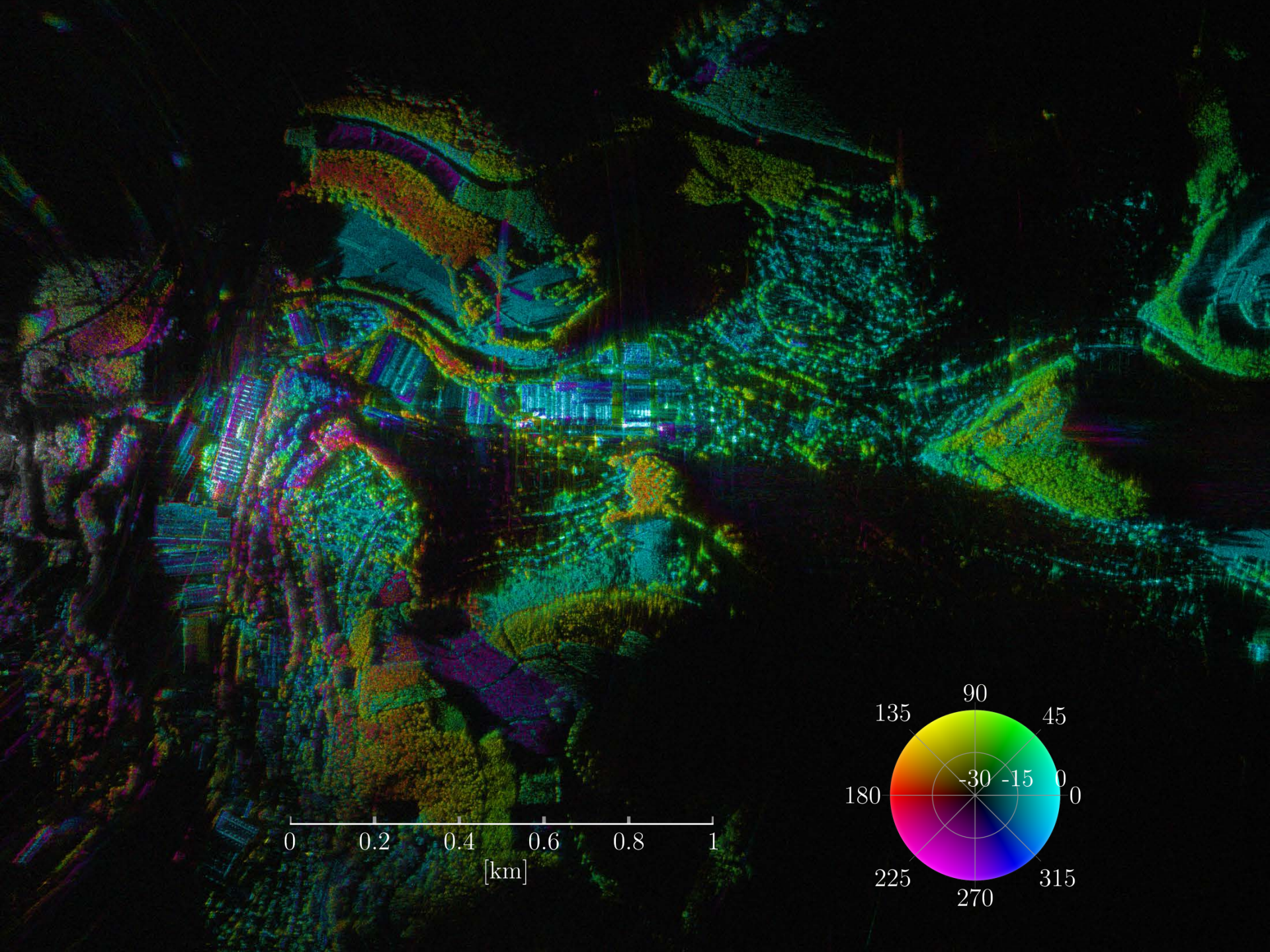


- Transmitter
- Target
- Receiver

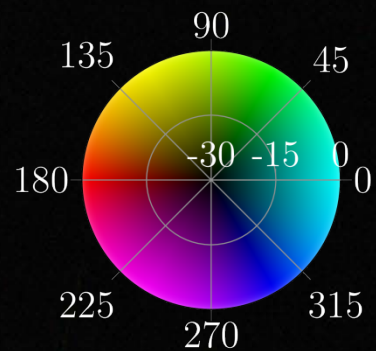
SAR Image Formation on GPU

Nvidia GeForce	GTX 670	GTX 980 Ti	GTX 1080
Radar Signal Size	147 MS	207 MS	?
SAR Image Size	209 MP	294 MP	?
Processing Speed	80 kP/s	160 kP/s	?
SP Compute Power	2460 GFlops	5632 GFlops	8228 GFlops
GPU Memory	2 GiB	6 GiB	8 GiB
Release Price / Year	\$400 / 2012	\$649 / 2015	\$599 / 2016

[1] F. Behner, S. Reuter, H. Nies and O. Loffeld (2016, Mar.). Synchronization and Processing in the HITCHHIKER Bistatic SAR Experiment. *IEEE Journal of Selected Topics in Applied Earth Observations and Remote Sensing*, vol. 9, no. 3, pp. 1028-1035.



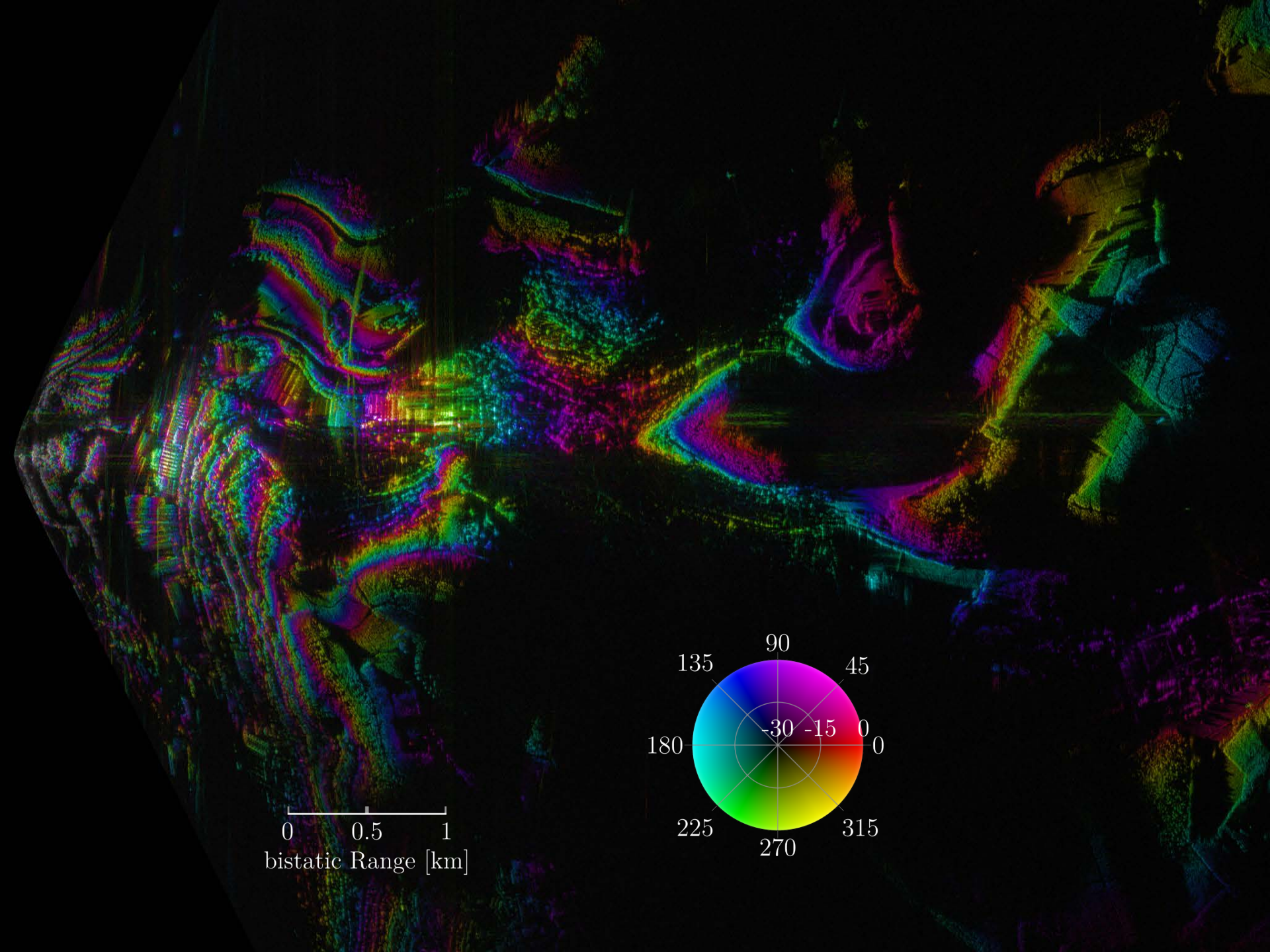
0 0.2 0.4 0.6 0.8 1
[km]



Bistatic Radar Coordinates

The bistatic radar coordinate system is a non-linear and surjective transformation of the position $\mathbf{p}$ in a Cartesian coordinate system and is defined by the zero-Doppler position X and the corresponding bistatic range R_b

$$X(\mathbf{p}) = \left\{ x : \frac{\partial R(x, \mathbf{p})}{\partial x} = 0 \right\}$$
$$R_b(\mathbf{p}) = R(X(\mathbf{p}), \mathbf{p})$$

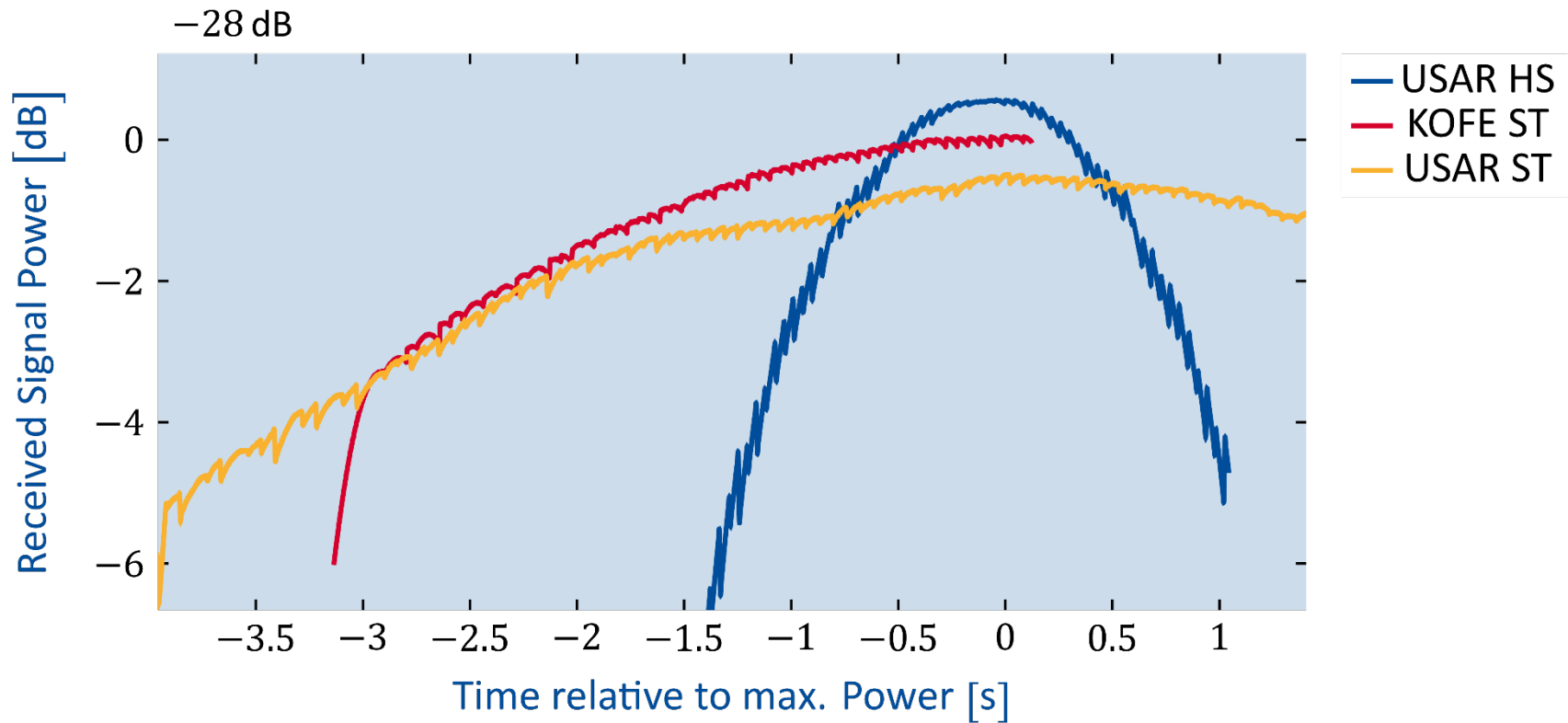


0 0.5 1
bistatic Range [km]



TerraSAR-X High Resolution Staring Spotlight Mode

A comparison of HS Mode / ST Mode



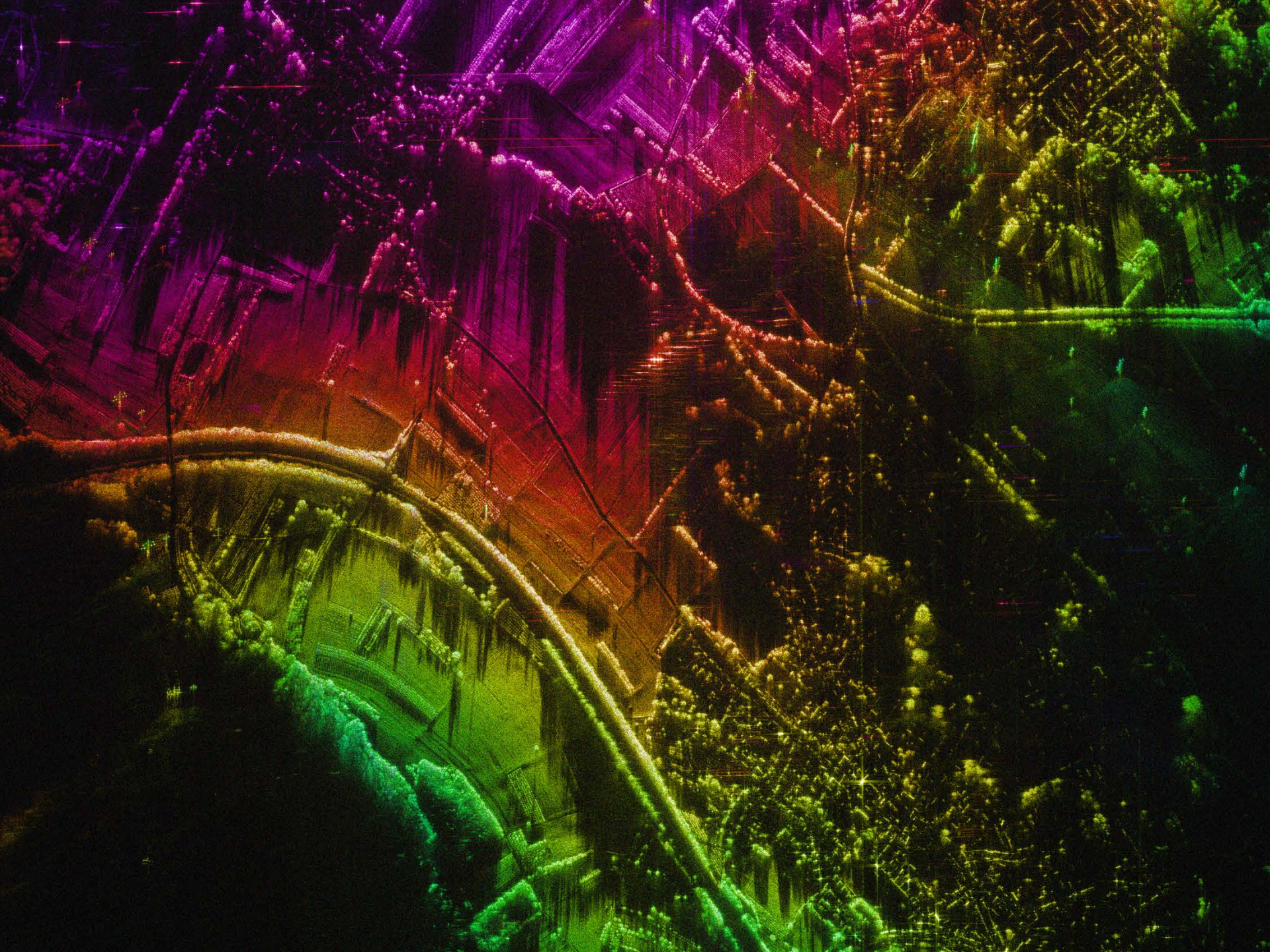
26.11.2014

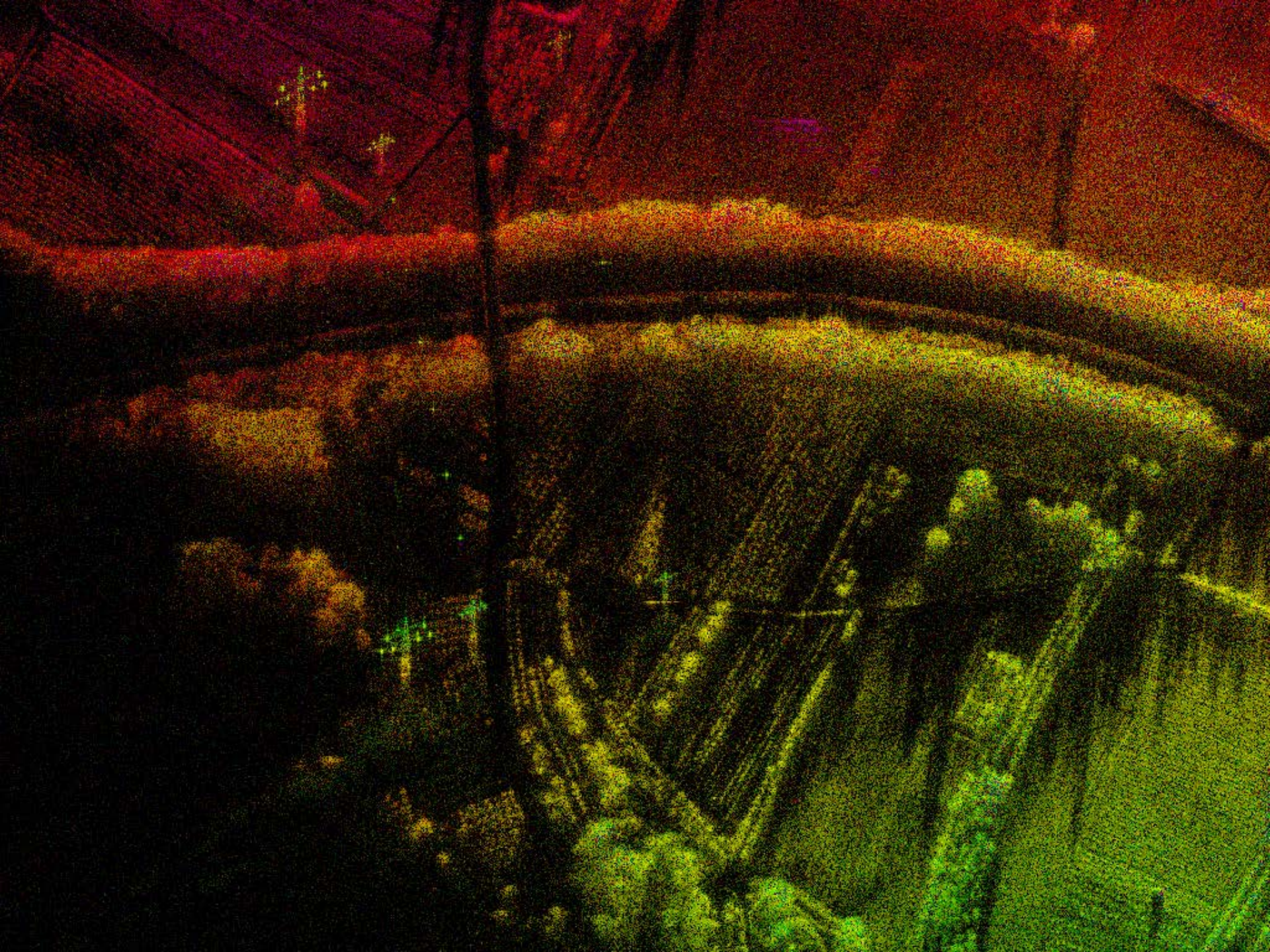
Along Track

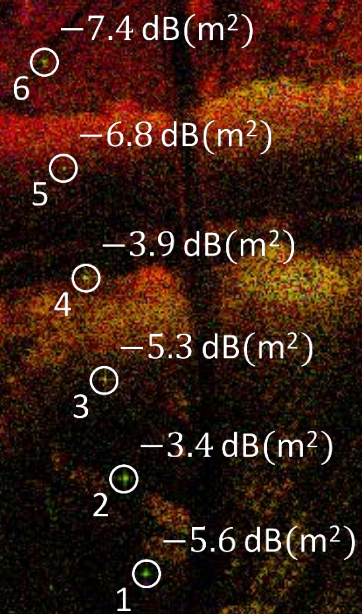
Range

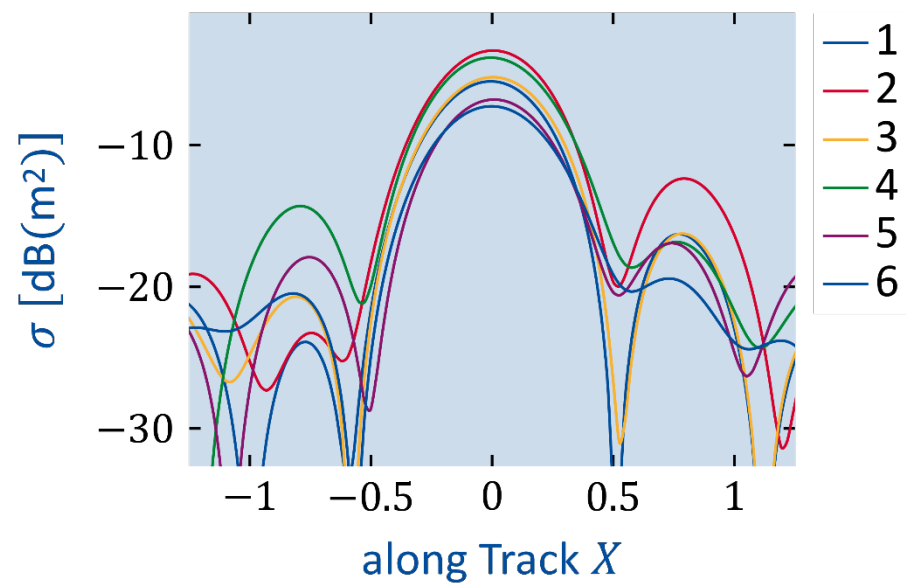
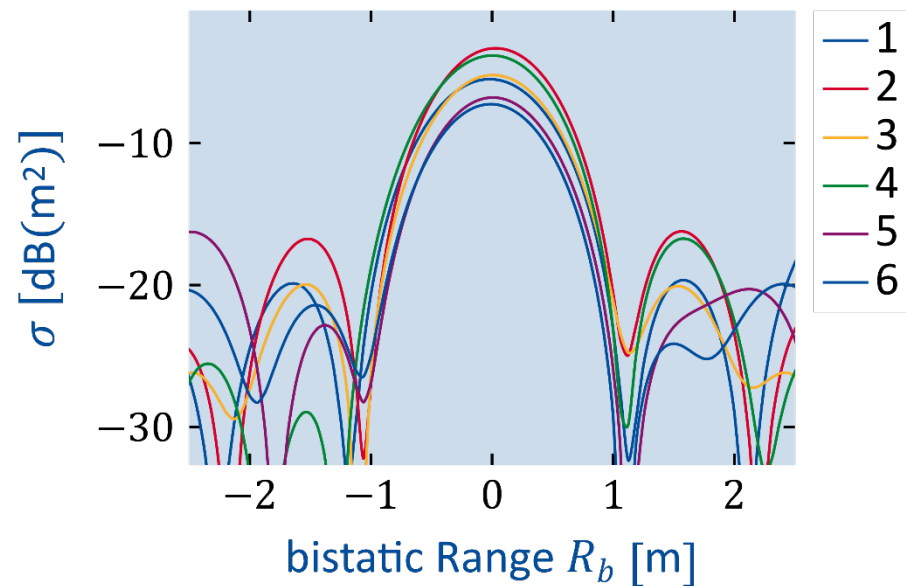
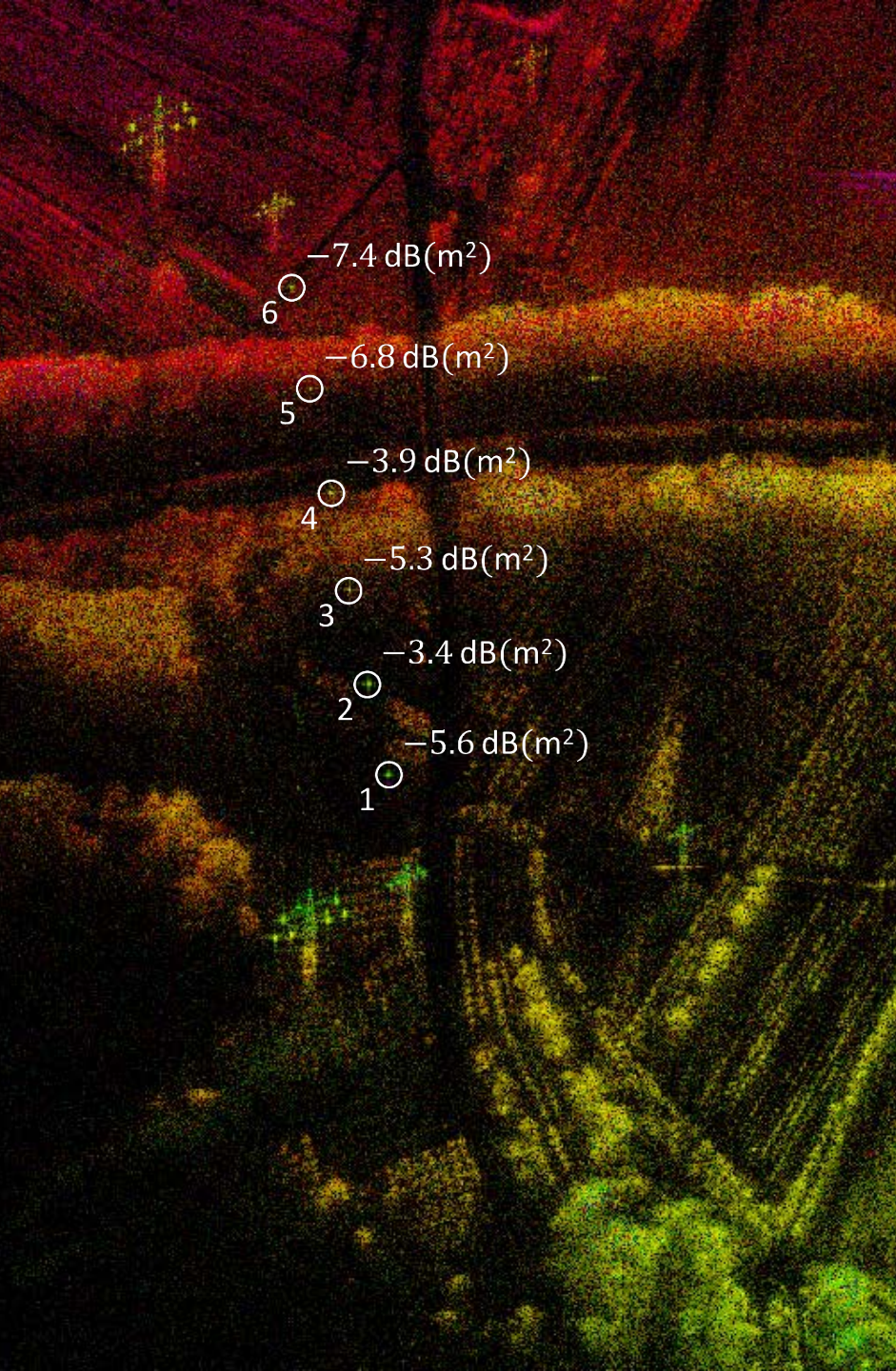
HITCHHIKER

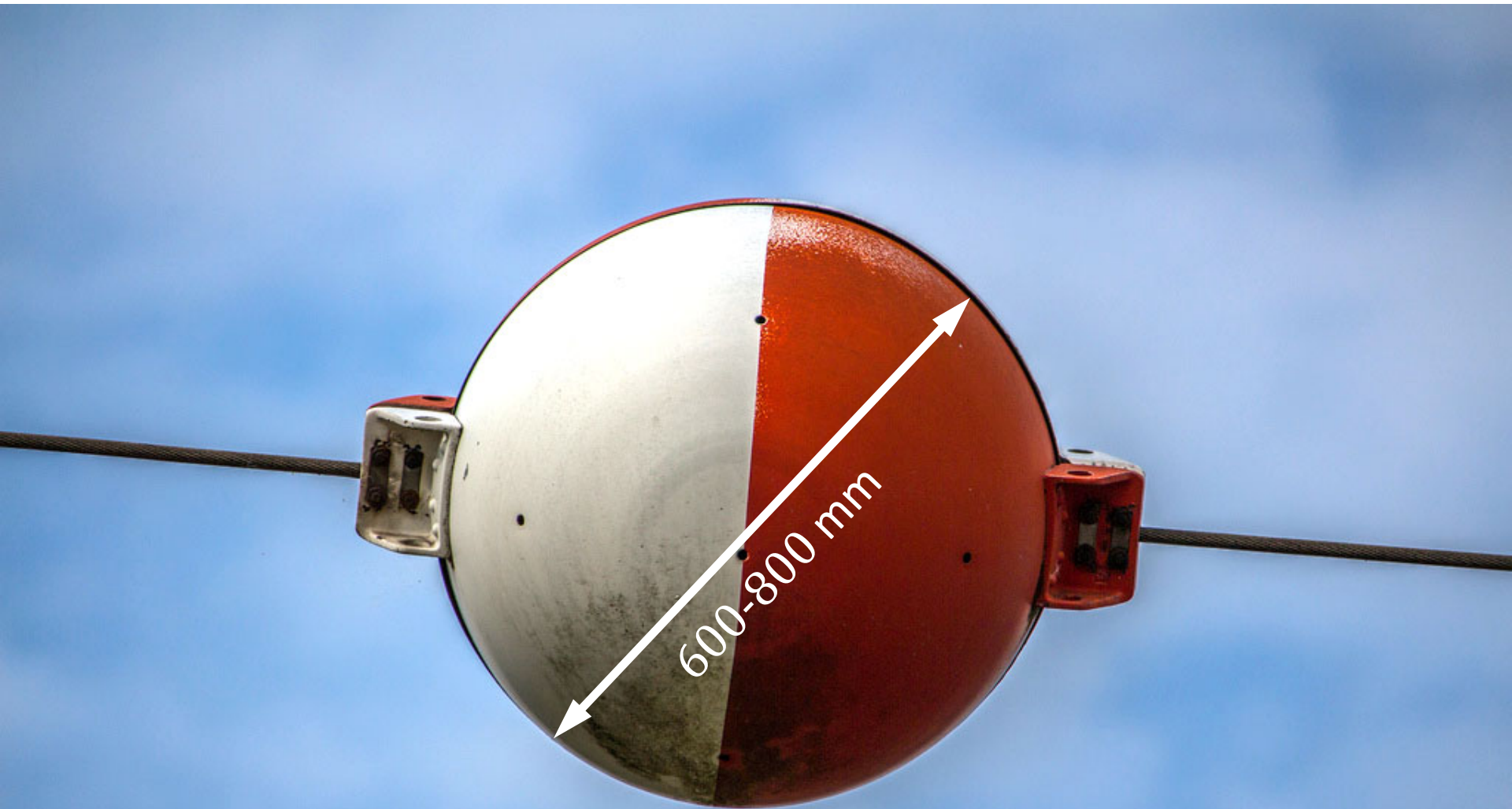






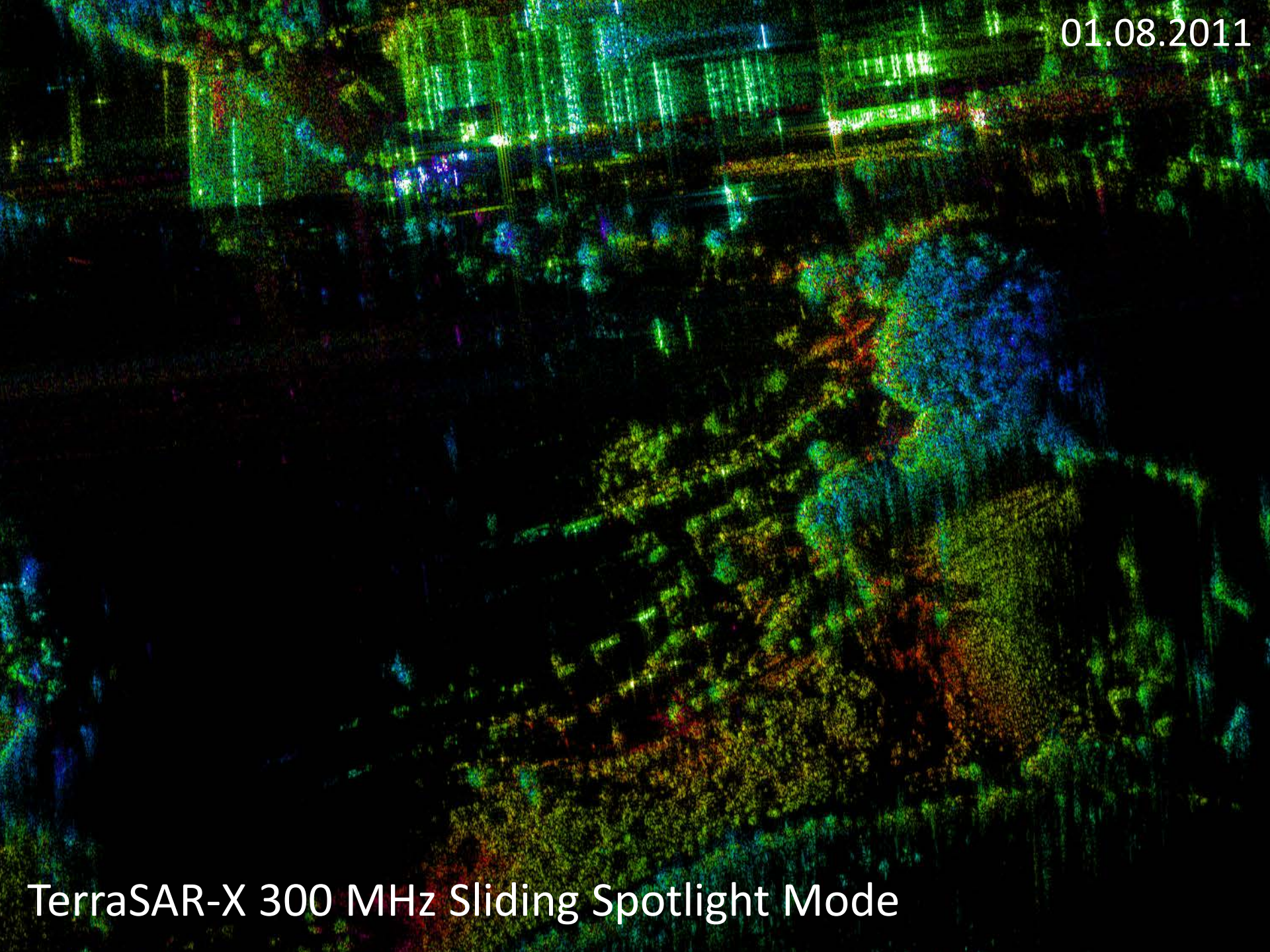






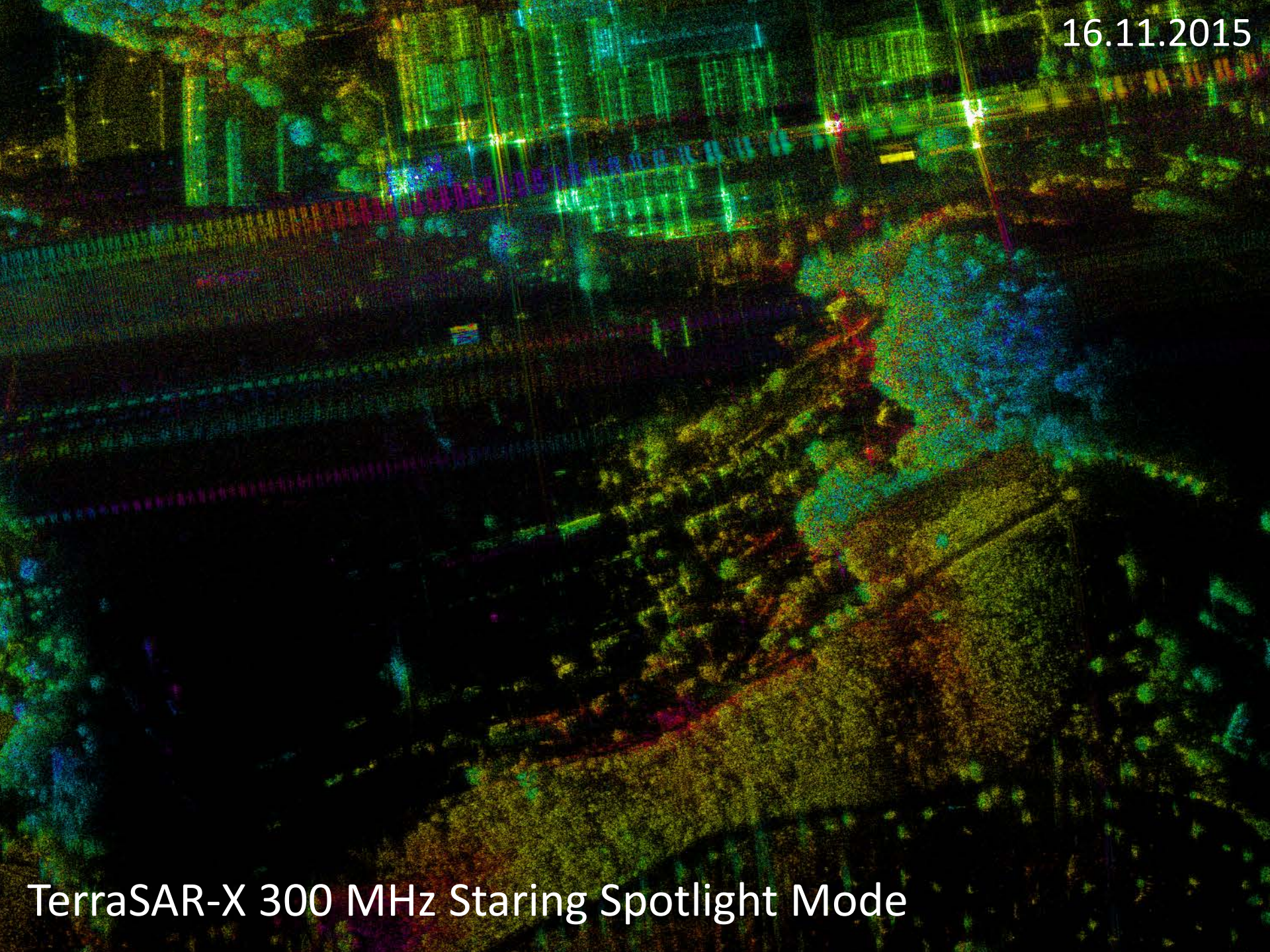
$$\sigma = -5.5 \text{ dB(m}^2\text{)} \dots -3 \text{ dB(m}^2\text{)}$$

01.08.2011



TerraSAR-X 300 MHz Sliding Spotlight Mode

16.11.2015



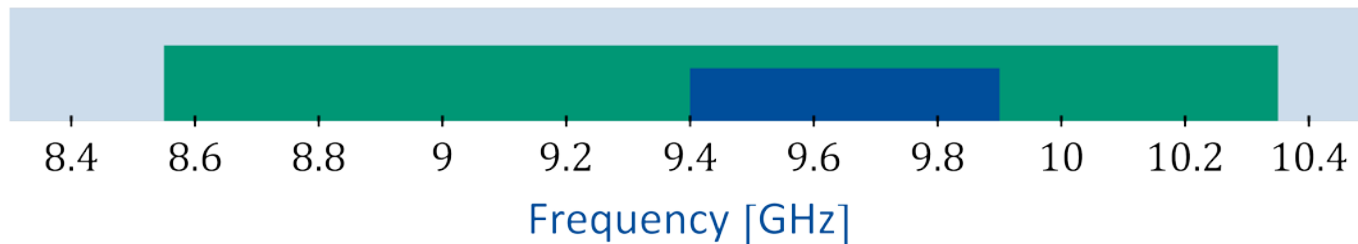
TerraSAR-X 300 MHz Staring Spotlight Mode

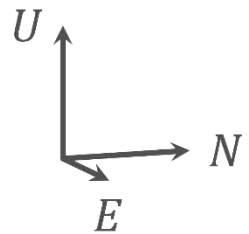
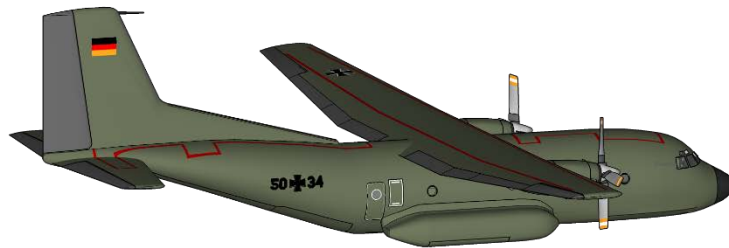
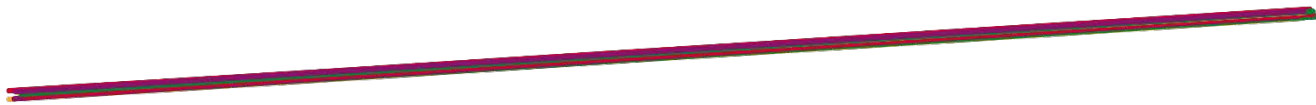
High Resolution Bistatic SAR Tomography Experiment



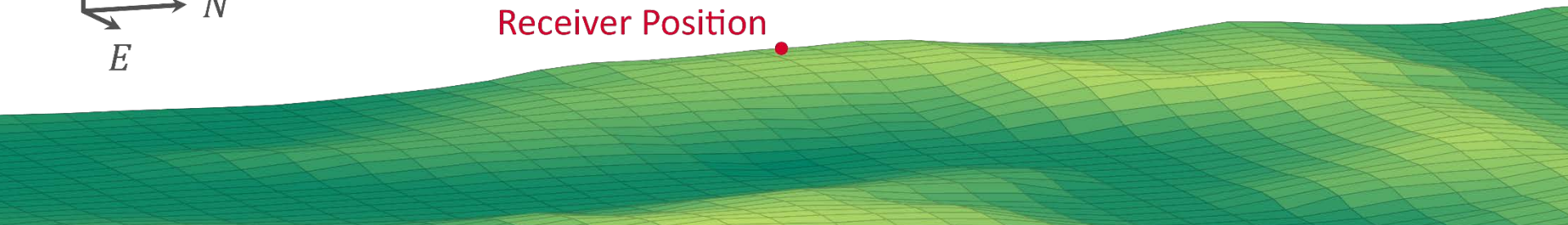
 **Fraunhofer**
FHR

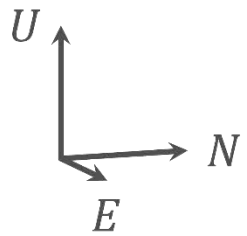
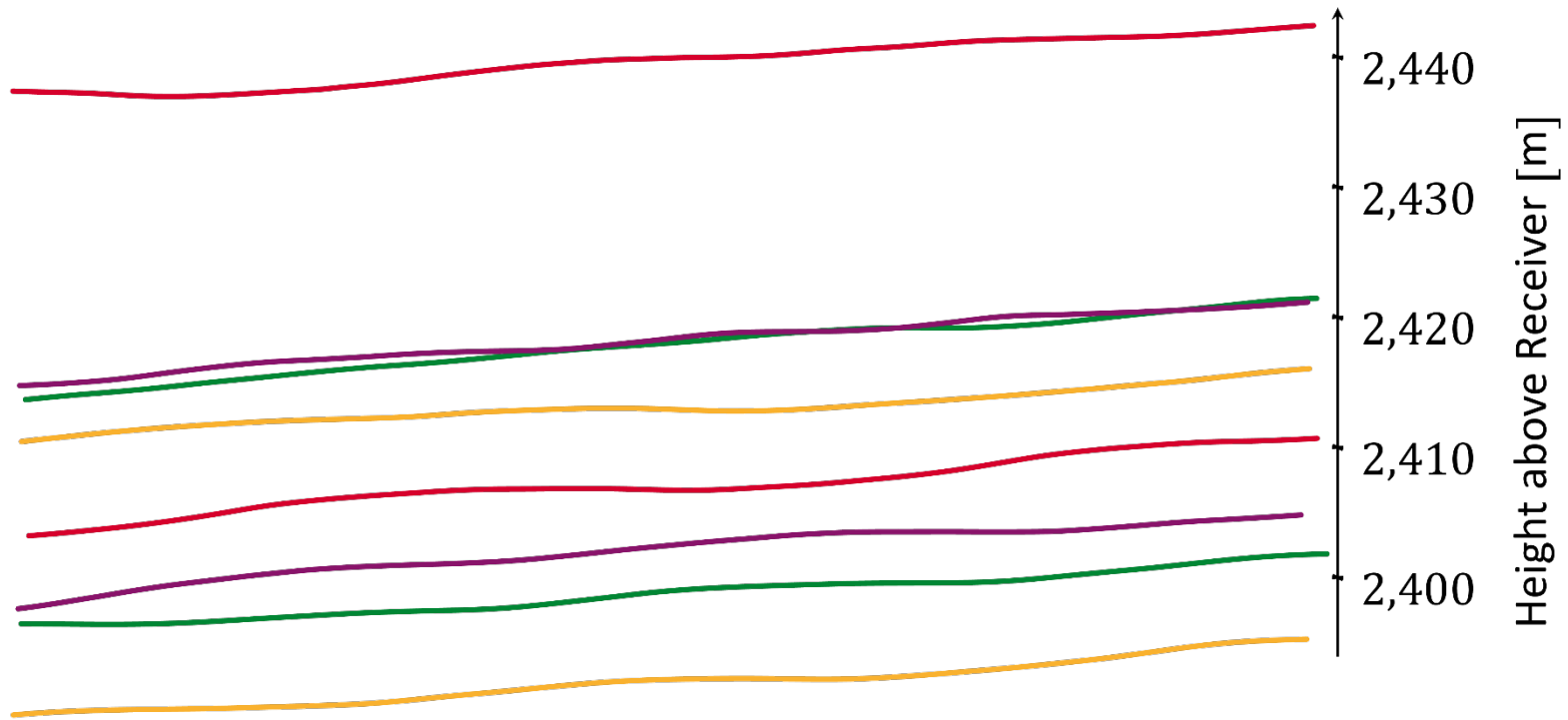
Bandwidth 1.8 GHz@9.45 GHz
Pulse Length 5.08 μ s
PRF 2 kHz
Mode Sliding Spotlight



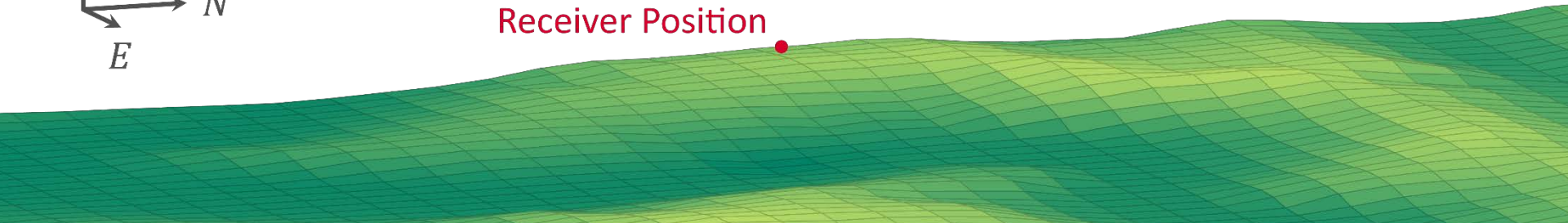


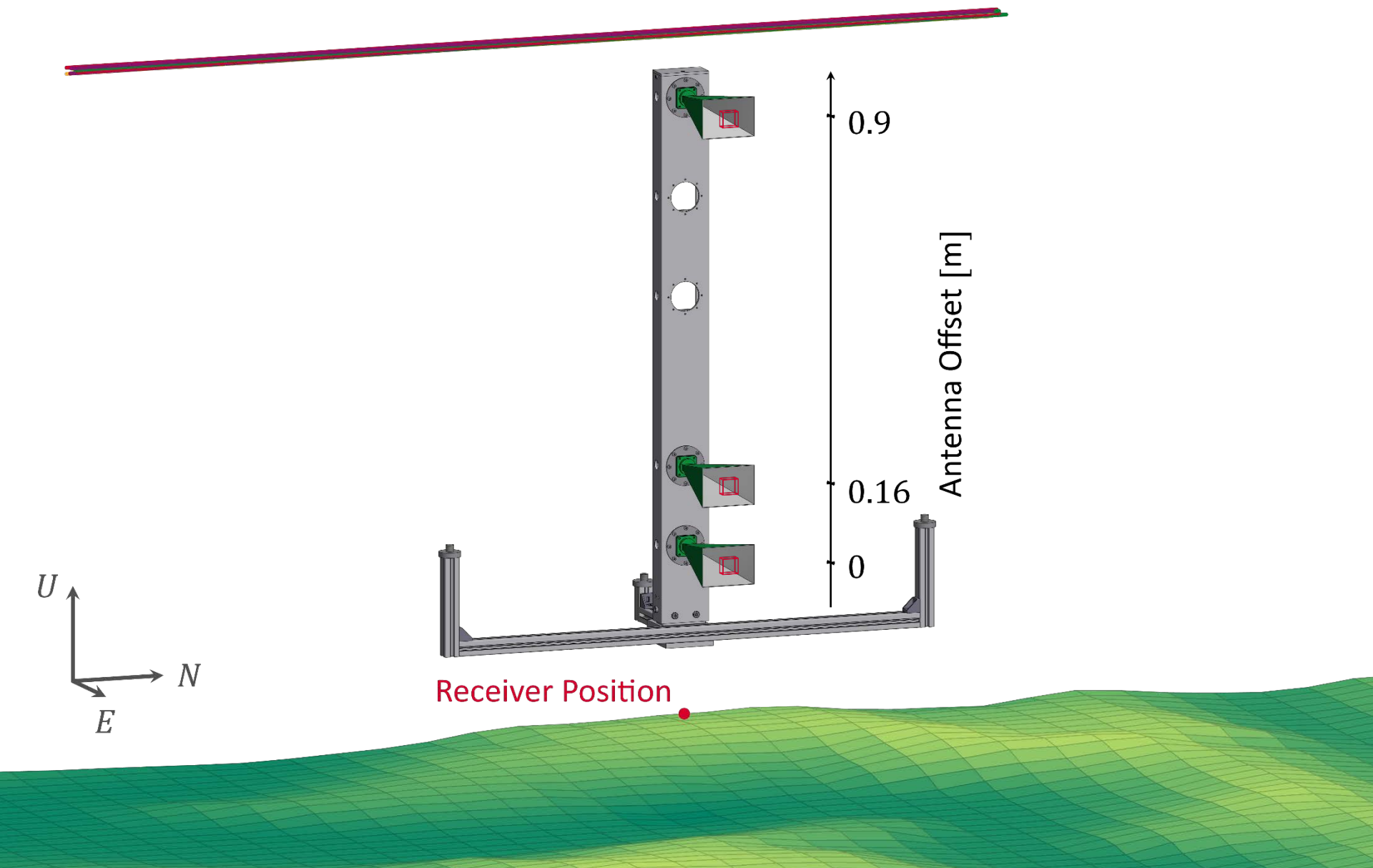
Receiver Position



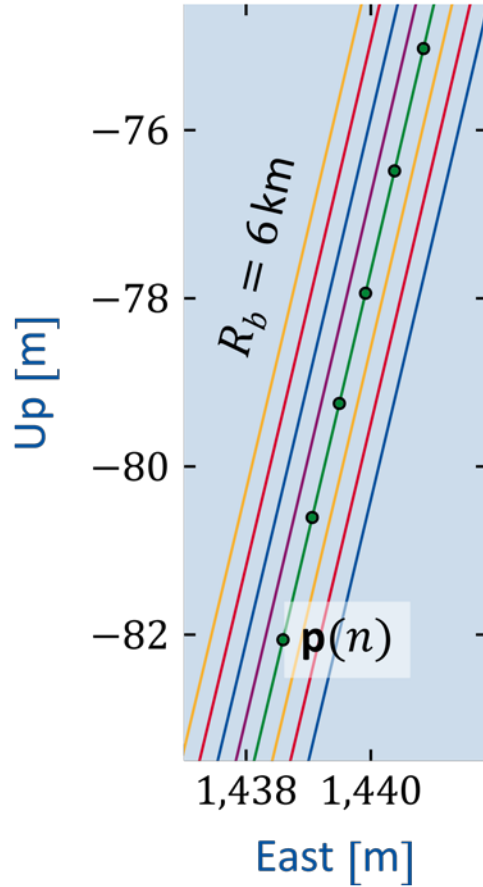


Receiver Position



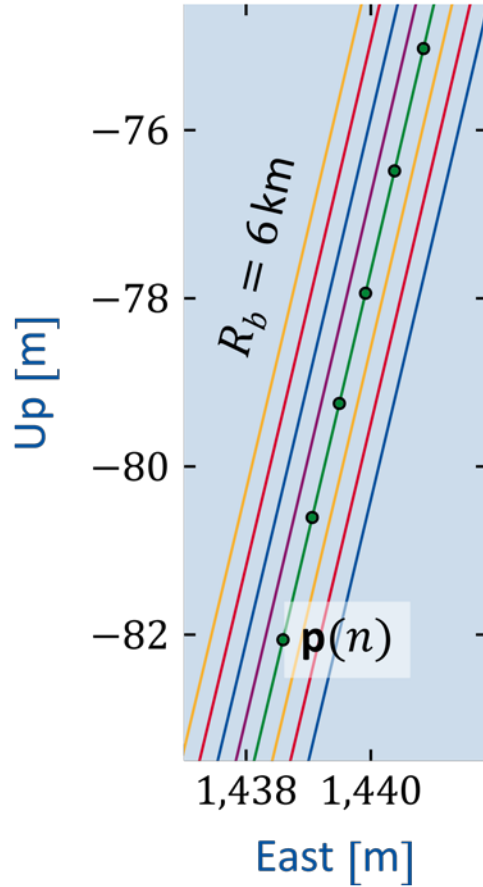


Tomo SAR Signal Model



Tomo SAR Signal Model

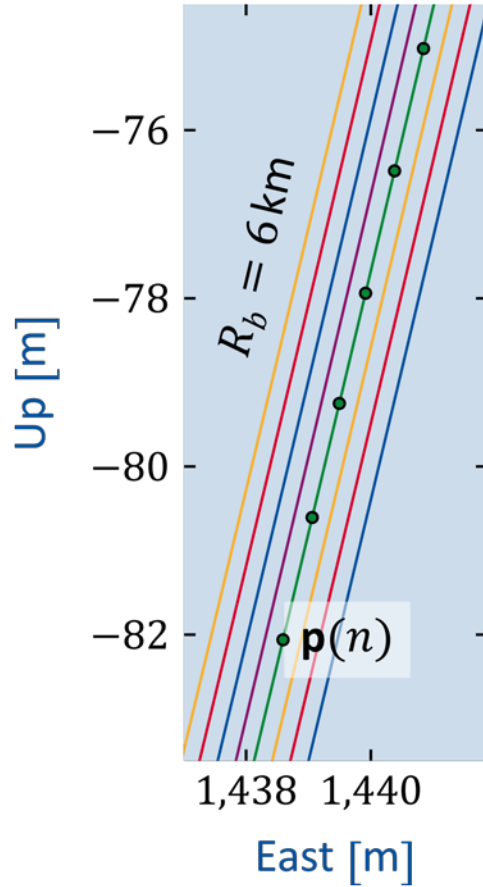
$$y = A \cdot x$$



Tomo SAR Signal Model

$$\mathbf{y} = \mathbf{A} \cdot \mathbf{x}$$

$$\mathbf{x} = (x_n) \in \mathbb{C}^N$$



Tomo SAR Signal Model

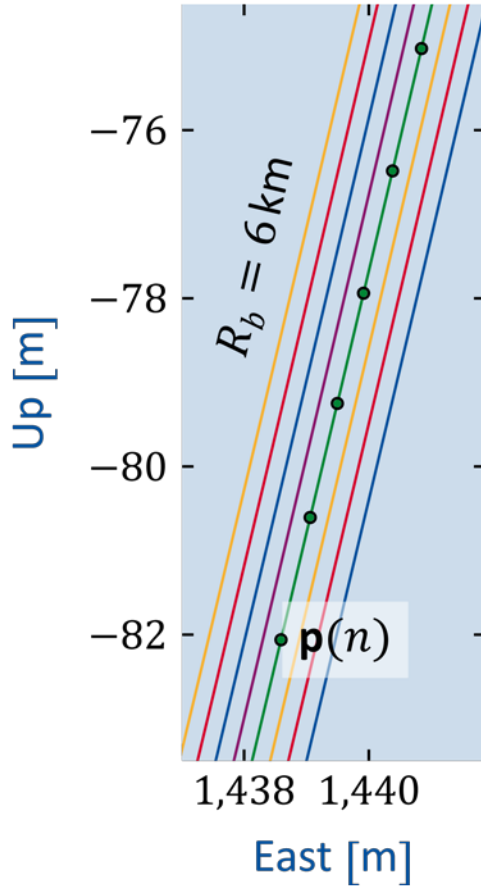
$$\mathbf{y} = \mathbf{A} \cdot \mathbf{x}$$

$$\mathbf{x} = (x_n) \in \mathbb{C}^N$$

$$R(x, m, n) = \|\mathbf{p}_t(x, m) - \mathbf{p}(n)\|_2 + \|\mathbf{p}_r(m) - \mathbf{p}(n)\|_2$$

$$X(m, n) = \left\{ x : \frac{\partial R(x, m, n)}{\partial x} = 0 \right\}$$

$$R_b(m, n) = R(X(m, n), m, n)$$



Tomo SAR Signal Model

$$\mathbf{y} = \mathbf{A} \cdot \mathbf{x}$$

$$\mathbf{x} = (x_n) \in \mathbb{C}^N$$

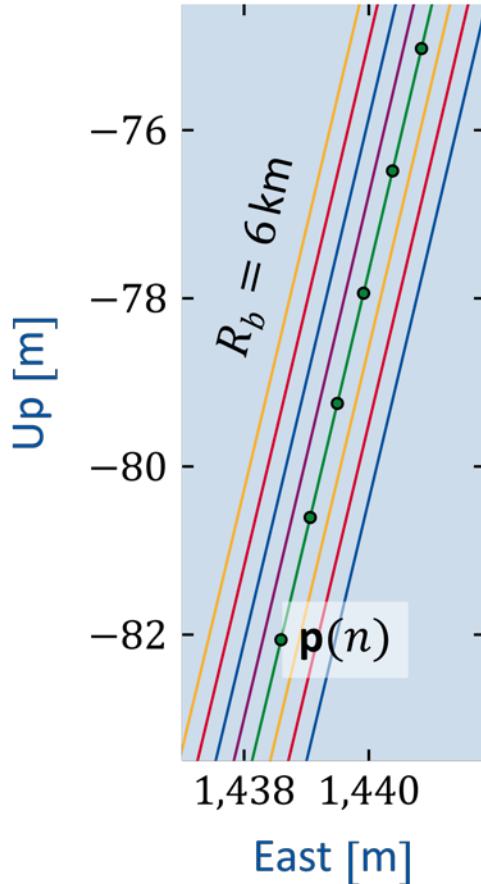
$$R(x, m, n) = \|\mathbf{p}_t(x, m) - \mathbf{p}(n)\|_2 + \|\mathbf{p}_r(m) - \mathbf{p}(n)\|_2$$

$$X(m, n) = \left\{ x : \frac{\partial R(x, m, n)}{\partial x} = 0 \right\}$$

$$R_b(m, n) = R(X(m, n), m, n)$$

$$\mathbf{A} = (a_{mn}) \in \mathbb{C}^{M \times N}$$

$$a_{mn} = \exp(-jk_0 R_b(m, n))$$



Tomo SAR Signal Model

$$\mathbf{y} = \mathbf{A} \cdot \mathbf{x}$$

$$\mathbf{x} = (x_n) \in \mathbb{C}^N$$

$$R(x, m, n) = \|\mathbf{p}_t(x, m) - \mathbf{p}(n)\|_2 + \|\mathbf{p}_r(m) - \mathbf{p}(n)\|_2$$

$$X(m, n) = \left\{ x : \frac{\partial R(x, m, n)}{\partial x} = 0 \right\}$$

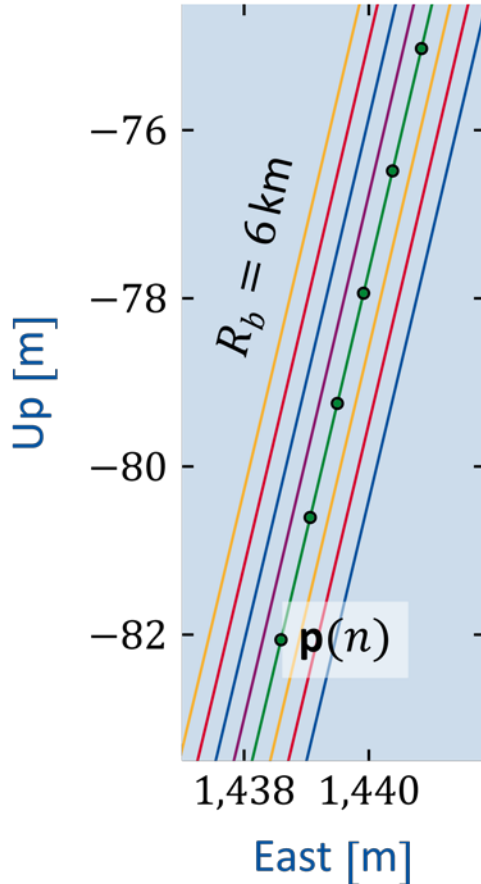
$$R_b(m, n) = R(X(m, n), m, n)$$

$$\mathbf{A} = (a_{mn}) \in \mathbb{C}^{M \times N}$$

$$a_{mn} = \exp(-jk_0 R_b(m, n))$$

$$\mathbf{y} = (y_m) \in \mathbb{C}^M$$

$$y_m = I_m(R_b(m, \cancel{n}), X(m, \cancel{n}))$$



Some definitions

Coherence of the sensing matrix:

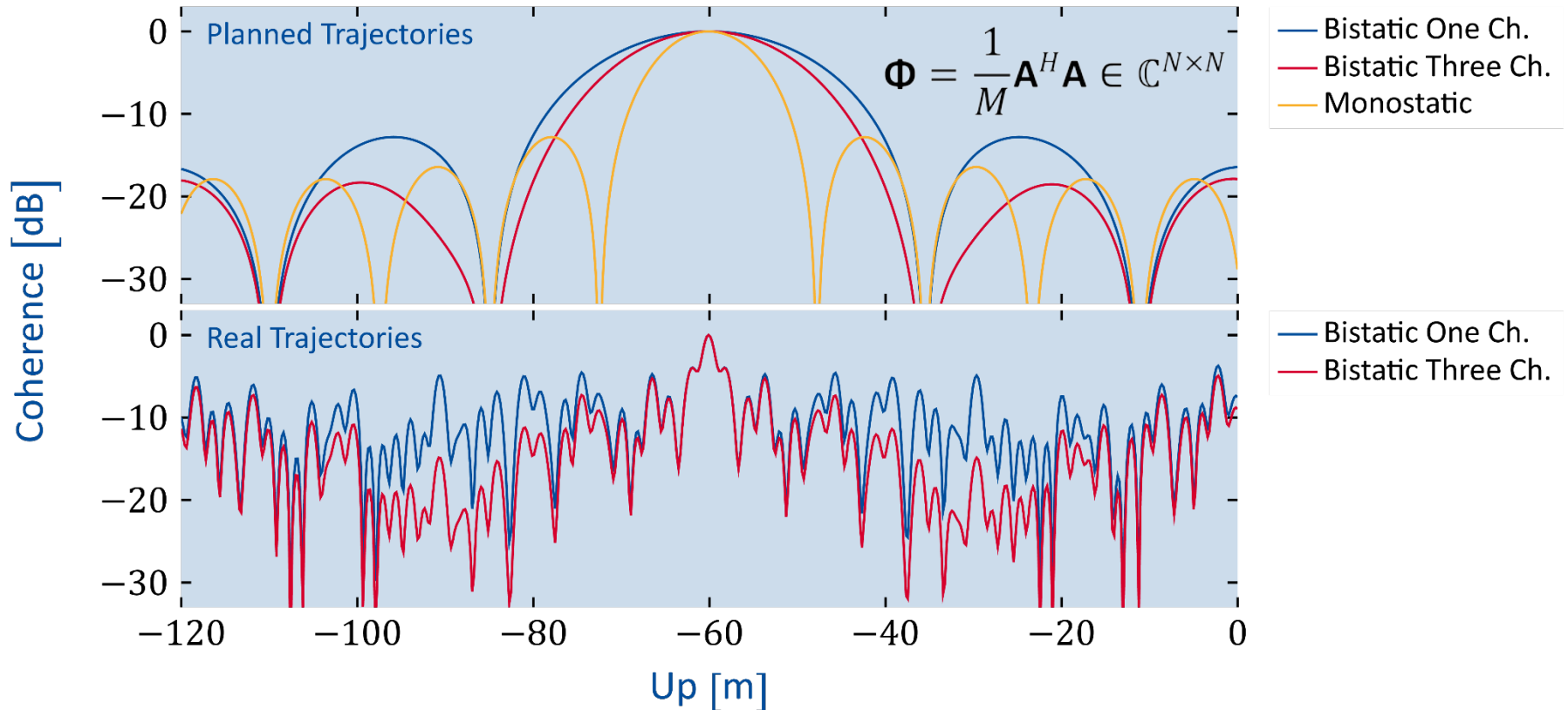
$$\mu(A) =: \max_{1 \leq i \neq j \leq N} \frac{|\langle \underline{a}_i, \underline{a}_j \rangle|}{\|\underline{a}_i\|_2 \cdot \|\underline{a}_j\|_2} = \max_{1 \leq i \neq j \leq N} \frac{1}{M} |\langle \underline{a}_i, \underline{a}_j \rangle|$$

Letting:

$$\Phi = \frac{1}{M} A^H A = \frac{1}{M} \begin{bmatrix} \langle \underline{a}_1, \underline{a}_1 \rangle & \langle \underline{a}_1, \underline{a}_2 \rangle & & \langle \underline{a}_1, \underline{a}_N \rangle \\ \langle \underline{a}_2, \underline{a}_1 \rangle & \langle \underline{a}_2, \underline{a}_2 \rangle & & \langle \underline{a}_2, \underline{a}_N \rangle \\ & & \ddots & \\ \langle \underline{a}_N, \underline{a}_1 \rangle & & & \langle \underline{a}_N, \underline{a}_N \rangle \end{bmatrix} = [\Phi_{i,j}] \quad \text{Matched filter result}$$

We have: $\Phi_{ii} = 1$ and $\Phi_{ij} = \frac{1}{M} \langle \underline{a}_i, \underline{a}_j \rangle$ and $\mu(A) = \max_{1 \leq i \neq j \leq N} |\Phi_{ij}|$

Plot of one row of: $\Phi_{\text{row } n_0} = \left[\frac{1}{M} \mathbf{A}^H \mathbf{A} \right]_{\text{row } n_0} = \left[\Phi_{n_0, j} \right]_{1 \leq j \leq N}$



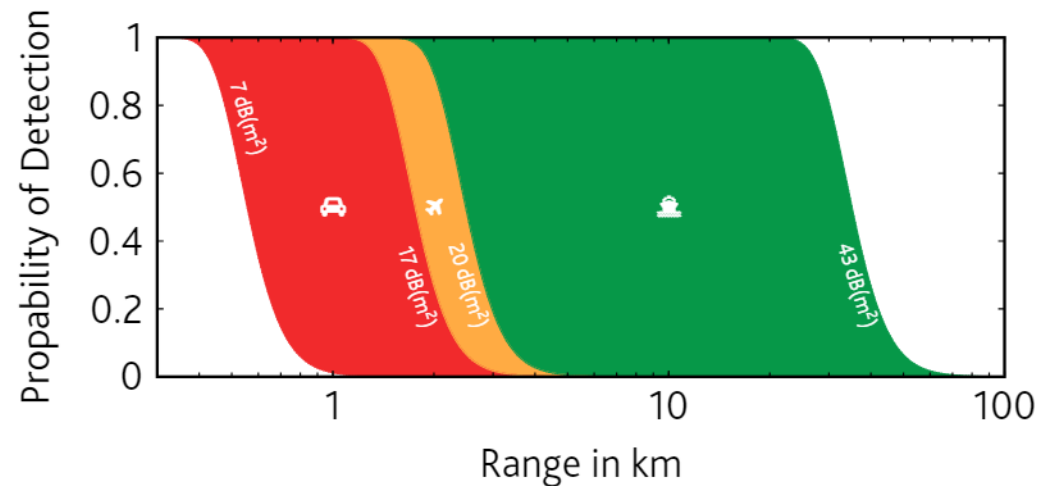
ℓ_1 - Regularized Least Squares

$$\hat{\mathbf{x}} = \arg \min_{\mathbf{x}} \left\{ \|\mathbf{x}\|_1 + \frac{\gamma}{2} \|\mathbf{Ax} - \mathbf{y}\|_2^2 \right\}$$

- [1] X. X. Zhu and R. Bamler (2012, Jan.). Super-Resolution Power and Robustness of Compressive Sensing for Spectral Estimation With Application to Spaceborne Tomographic SAR. *IEEE Transactions on Geoscience and Remote Sensing*, vol. 50, no. 1, pp.247-258.
- [2] S. Foucart and H. Rauhut, "Chambolle and Pock's Primal-Dual Algorithm", in *A Mathematical Introduction to Compressive Sensing*, Springer New York, 2013, pp. 486-487.

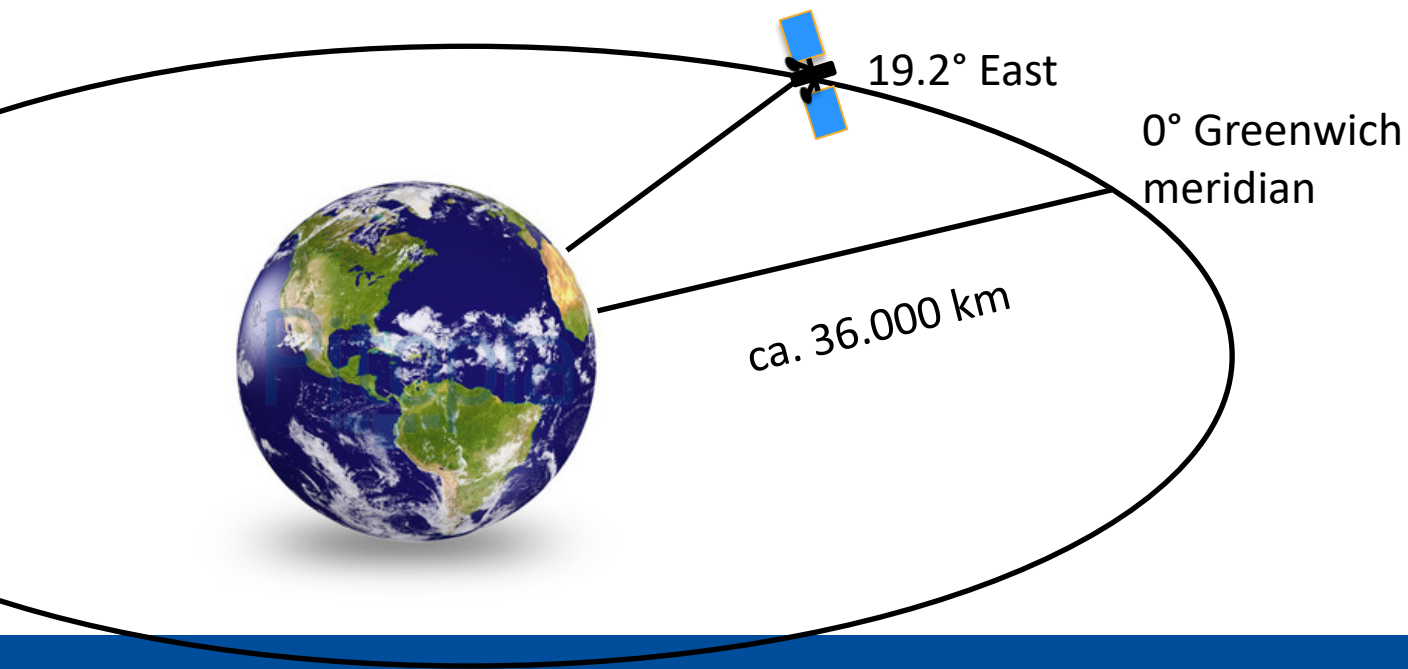
Radar processing using geostationary satellites

- Permanent availability
- Cost effective
- Large coverage
- Different Applications
 - Imaging
 - Change detection
 - Tracking



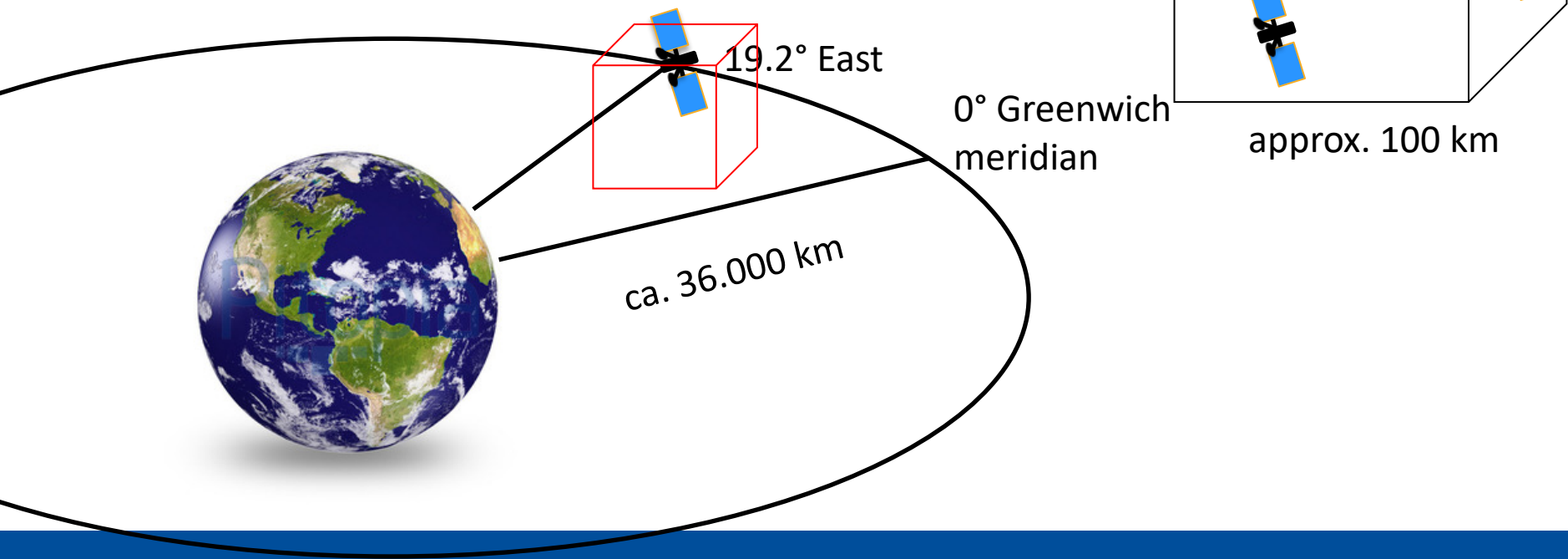
Astra Satellite System

- Geostationary Orbit
- Co-located Satellites
- Television and Radio Broadcasting

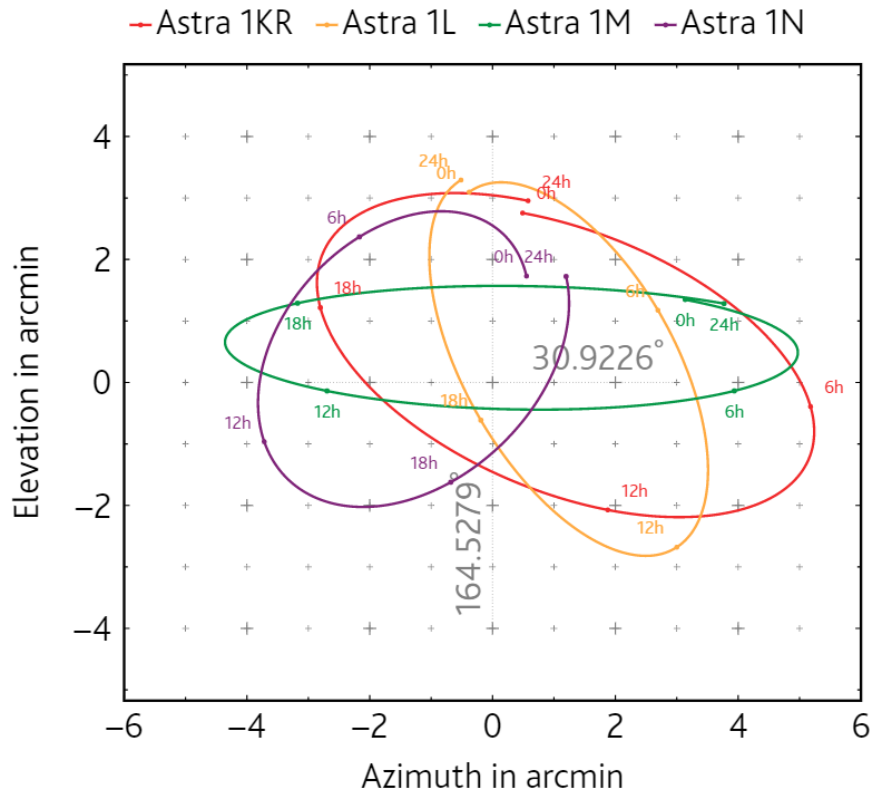


Astra Satellite System

- Geostationary Orbit
- Co-located Satellites
- Television and Radio Broadcasting

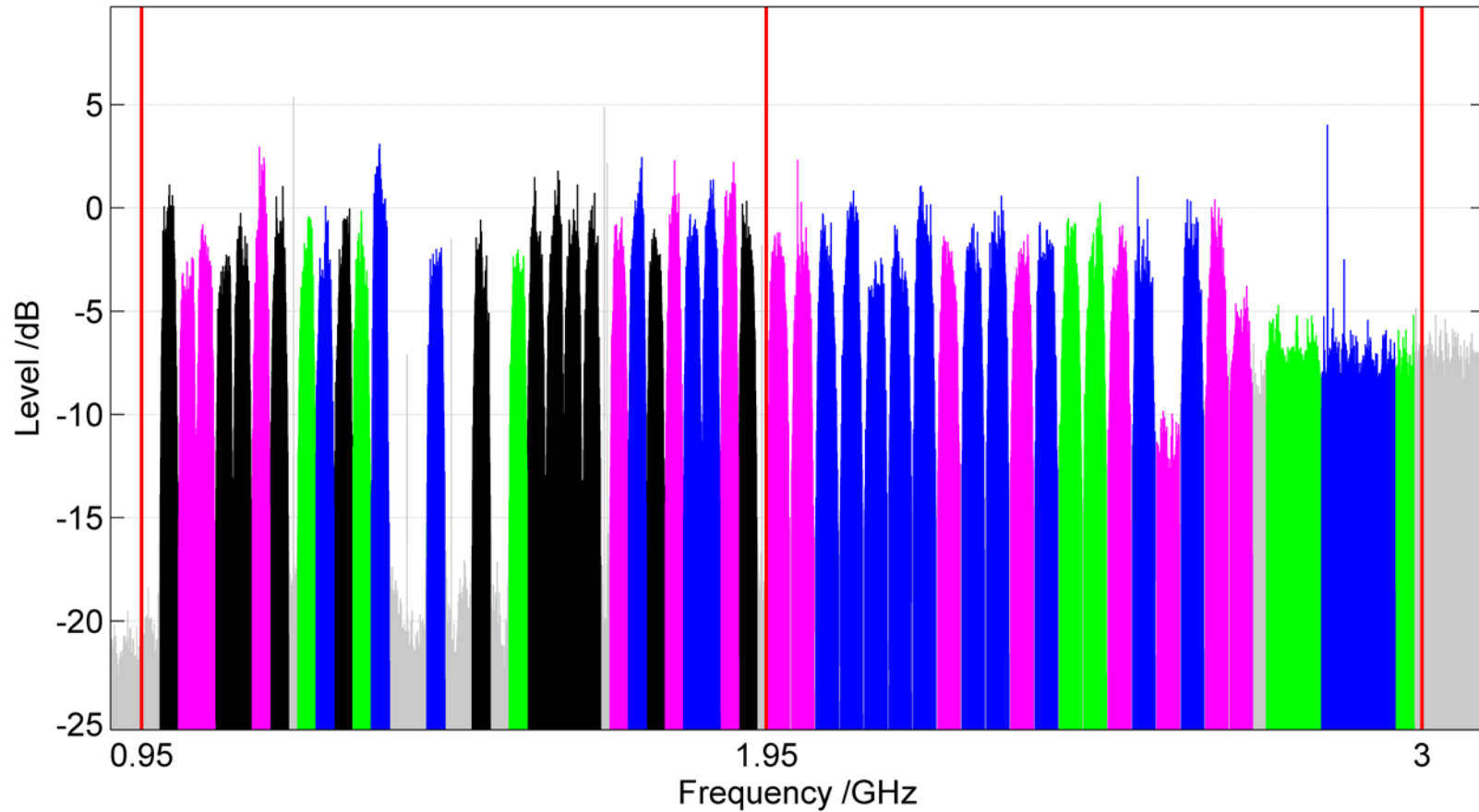


Astra Satellite System



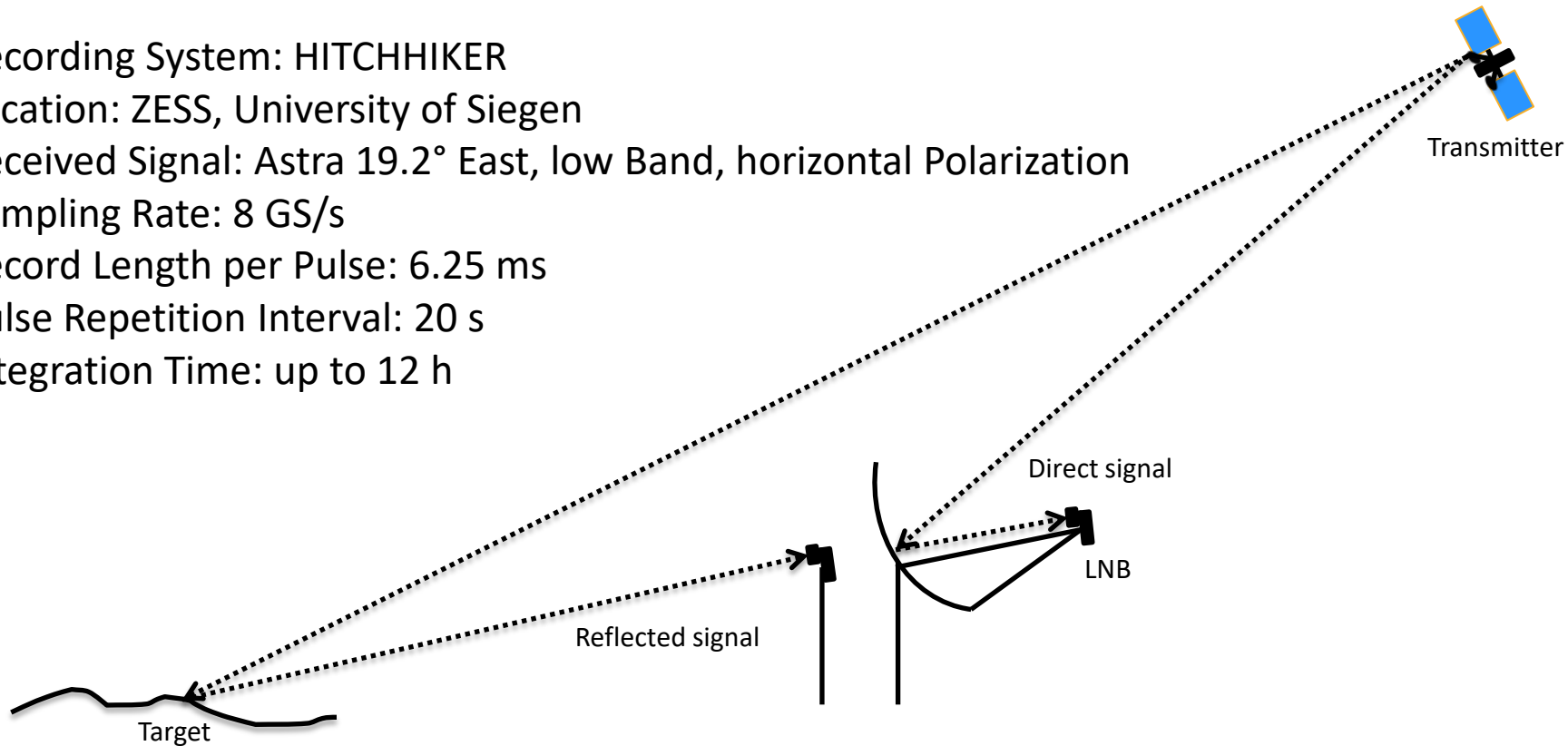
- Ku-Band 10.7 GHz – 12.75 GHz
- Transponders distributed on different Satellites
- Cross Range Resolution (SAR): 13 m
- Range Resolution:
 - Full Bandwidth (2.05 GHz): 15 cm
 - 1 Transponder (30 MHz): 10 m

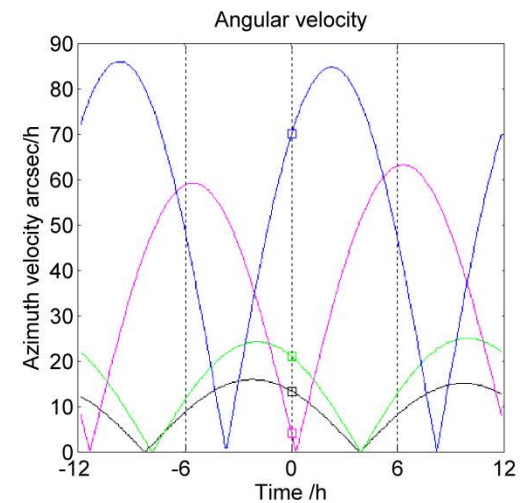
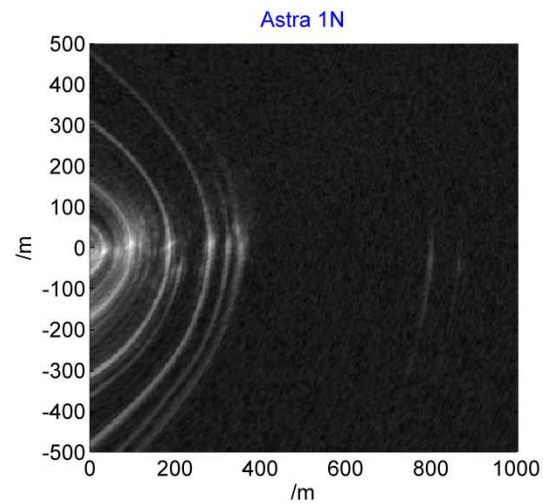
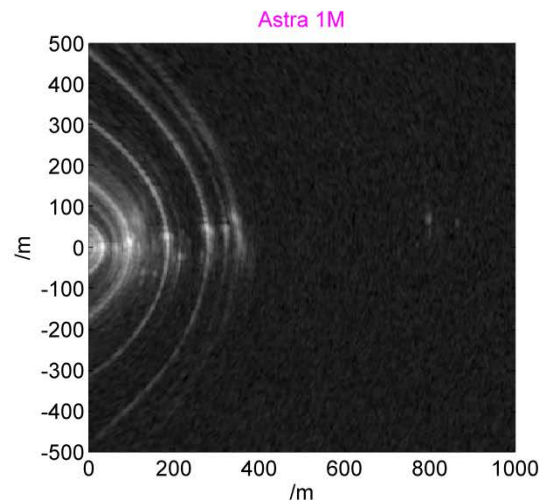
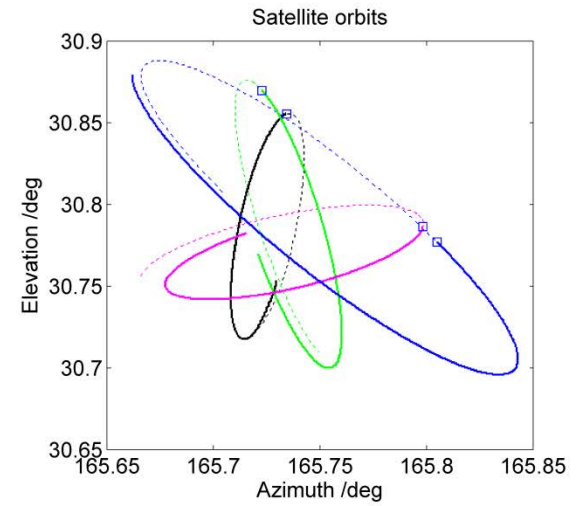
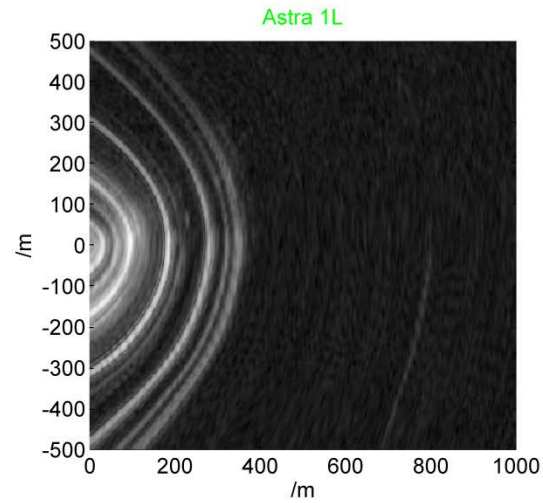
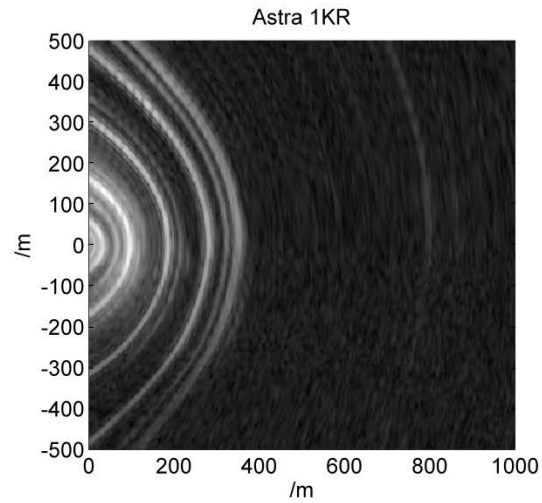
Astra spectrum, transponders of Astra 1KR, Astra 1L, Astra 1M, Astra 1N



SAR Imaging Experiment

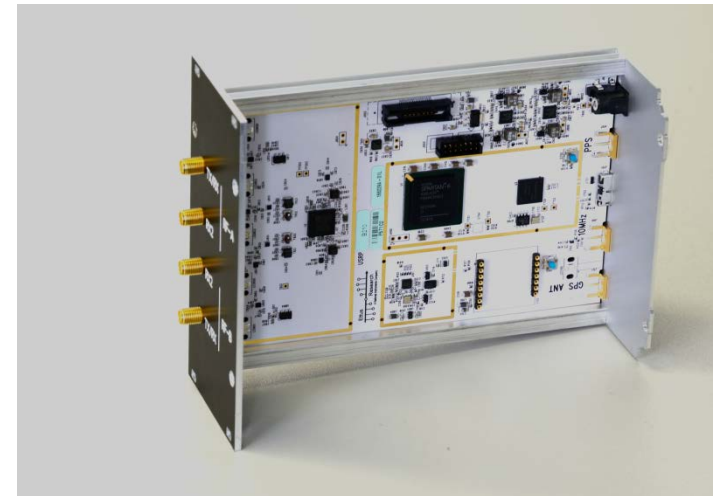
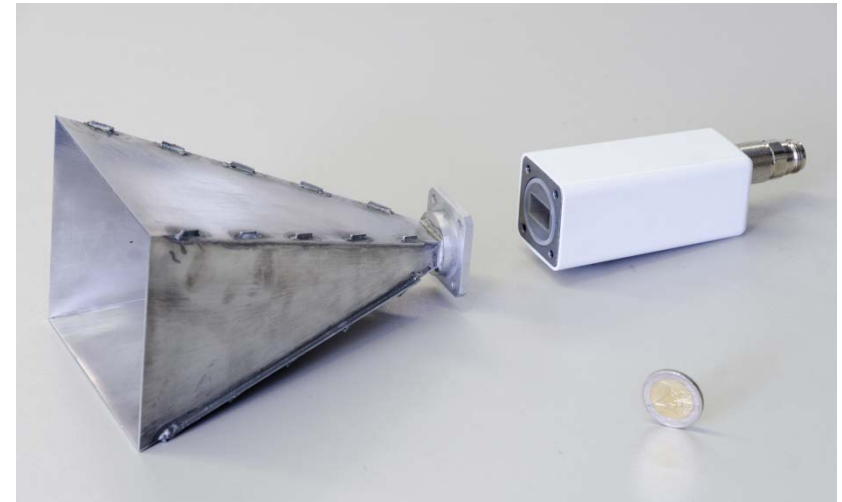
- Recording System: HITCHHIKER
- Location: ZESS, University of Siegen
- Received Signal: Astra 19.2° East, low Band, horizontal Polarization
- Sampling Rate: 8 GS/s
- Record Length per Pulse: 6.25 ms
- Pulse Repetition Interval: 20 s
- Integration Time: up to 12 h





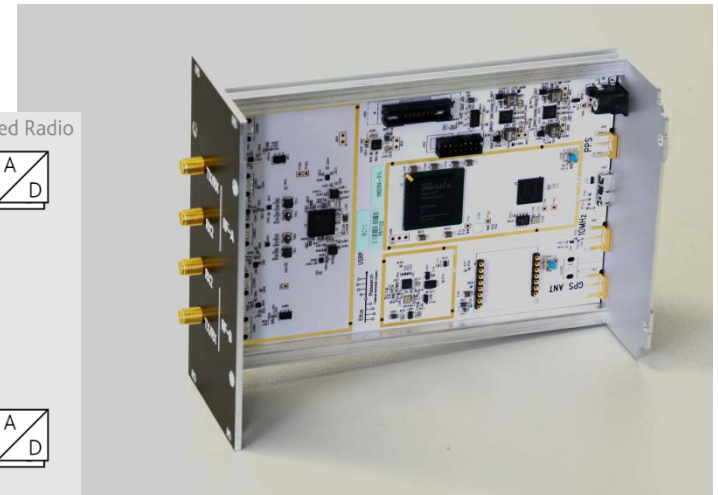
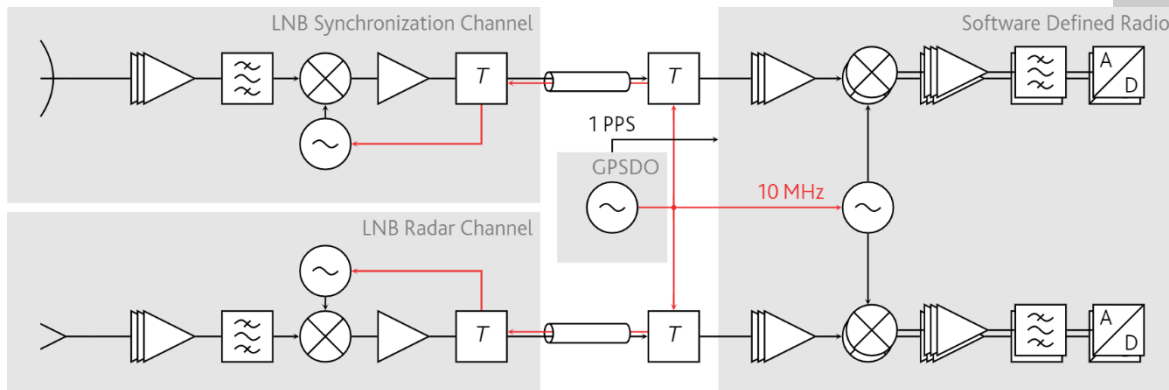
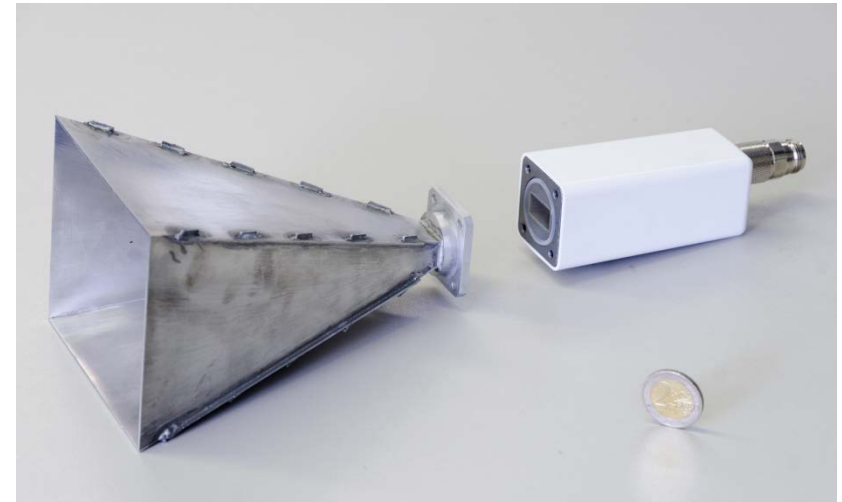
Passive Radar System

- Dedicated DVB-S passive radar
- Radar applications
 - Moving target detection/Tracking
 - Moving receiver platform
 - ISAR imaging
 - Modular multistatic system



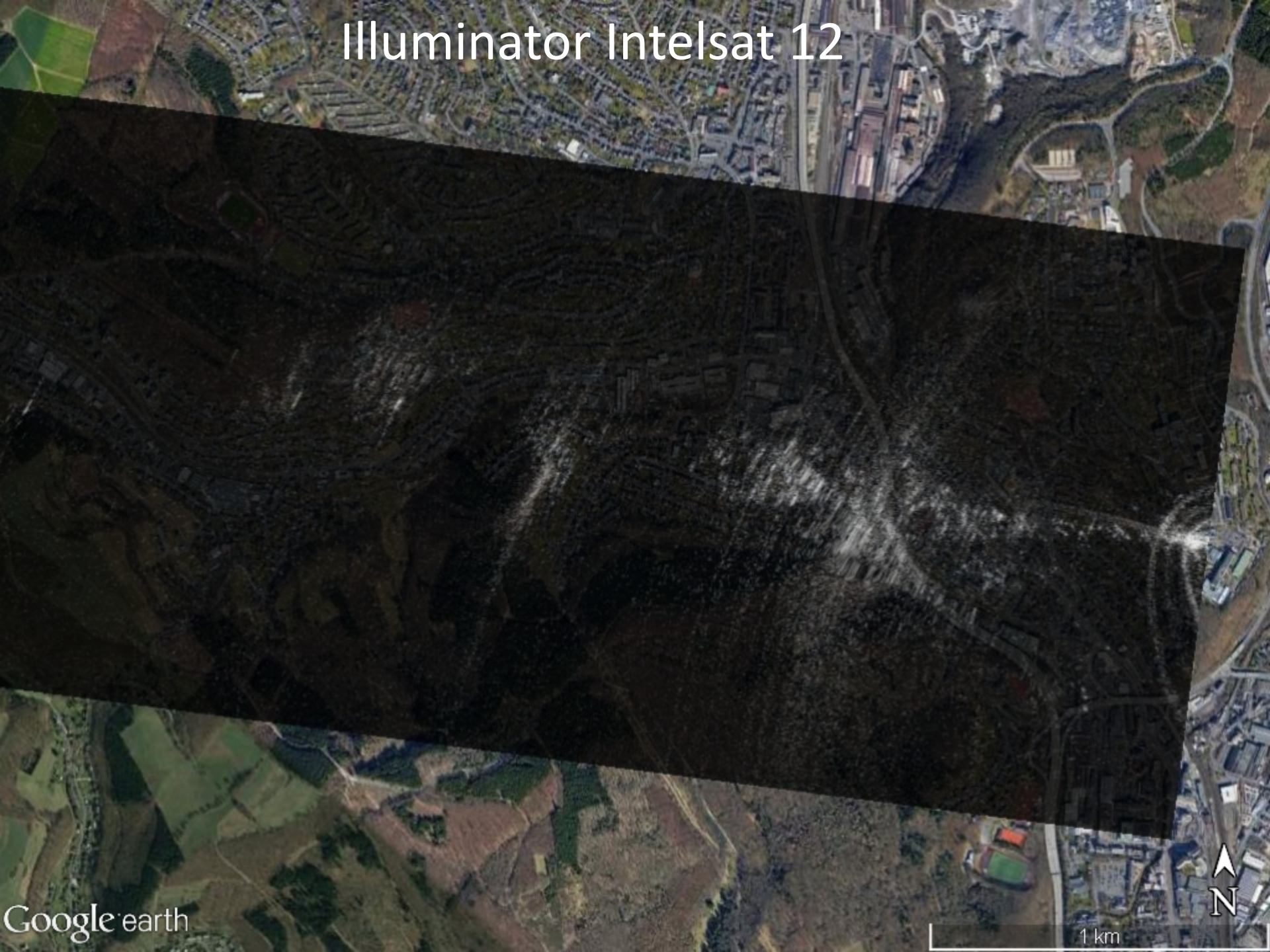
Passive Radar System

- Dedicated DVB-S passive radar
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Illuminator Intelsat 12



The (preliminary) end, but...



CoSeRa 2018

5th International Workshop on Compressed Sensing
applied to Radar, Multimodal Sensing and Imaging

10.-13. September 2018 – University of Siegen

www.zess.uni-siegen.de/cosera2018



Thank you for your attention!

